TF 2300

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F.M./A.M. Modulation Meter

INSTRUCTION MANUAL

MP MASTER
MA23645

MARCONI INSTRUMENTS LIMITED
ST. ALBANS HERTFORDSHIRE ENGLAND

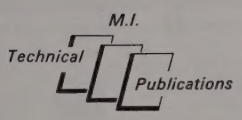
INSTRUCTION MANUAL

No. EB 2300

for

F.M./A.M. Modulation Meter

TF 2300



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ST. ALBANS HERTFORDSHIRE ENGLAND

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Contents

Section 1 GENERAL INFORMATION

1.1 Introduction 3
1.2 Data summary 4
1.3 Accessories 5

Section 2 OPERATION

2.1 Installation 7
2.2 Power supply 7
2.3 Controls 8
2.4 Preparation for use 10
2.5 Measuring f. m. deviation ... 10
2.6 Measuring a. m. depth ... 11
2.7 Measuring f. m. on a. m. ... 11
2.8 Measuring a. m. on f. m. ... 12
2.9 Noise measurements 12
2.10 Oscillator arrangements and operating notes 14
2.11 Measurement in 1 to 2 Mc/s range 14
2.12 Asymmetric modulation and carrier shift 15
2.13 Use of l. f. output terminals ... 15
2.14 Use of i. f. output socket ... 15
2.15 Crystal selection 15
2.16 F. M. stereo measurements ... 16

Decibel conversion table ... 20

Section 3 TECHNICAL DESCRIPTION

3.1 System operation 23
3.2 Power unit 24
3.3 Mixer 25
3.4 Local oscillator 25
3.5 I. F. amplifier 26
3.6 Limiter 26
3.7 Discriminator 26
3.8 Calibrator 26
3.9 Low-pass filters 27
3.10 1st l. f. amplifier 27
3.11 2nd l. f. amplifier 27
3.12 Peak reading meter 28
3.13 A. M. detector 28
3.14 Between-units circuitry on chassis 28

Section 4 MAINTENANCE

4.1 Introduction 29
4.2 Fuse replacement 29
4.3 Access and layout 29
4.4 Cleaning and lubricating ... 30
4.5 Replacing oscillator drive cord 30

Component layout illustrations 31

4.6 Checking meter calibration ... 33

Section 5 (*not assigned*)

Section 6 REPLACEABLE PARTS

Introduction 35

Section 7 CIRCUIT DIAGRAMS

Circuit notes 48

Fig. 7.1 Inter-unit wiring 49
Fig. 7.2 Mixer, oscillator and i. f. amplifier 51
Fig. 7.3 Limiter, discriminator and 1st l. f. amplifier 53
Fig. 7.4 2nd l. f. amplifier and voltmeter 55
Fig. 7.5 A. M. detector and calibrator 57
Fig. 7.6 Power supply unit 59

Section
I

General information

1.1 INTRODUCTION

The TF 2300 Modulation Meter is primarily for measurement of f. m. deviation but it also measures a. m. depth. With its wide range of deviation frequency, modulation bandwidth and carrier frequency, this instrument is suitable for application to fixed and mobile point-to-point communications, broadcasting, telemetry and multi-channel link equipment in the h. f. , v. h. f. and u. h. f. bands. Distortion and channel separation tests on f. m. stereo receivers and transmitters can also be made.

Positive and negative f. m. deviation can be measured in ranges from 5 kc/s to 500 kc/s full-scale at modulation frequencies between 30 c/s and 150 kc/s. A. M. depth can be measured up to 95% in a 30 c/s to 15 kc/s modulation bandwidth. Either f. m. or a. m. can be measured in the presence of the other. Although measurements are normally made by means of the meter readout, i. f. and demodulated outputs are available at the front panel for examination or analysis.

Spurious a. m. and f. m. due to hum and noise are kept to a level insignificant for most applications but, where required, crystals can be switched in to control the local oscillator, or an external local oscillator may be used. The instrument can be operated without a local oscillator for measurements in the 1-2 Mc/s range.

The instrument can be operated from mains power or a nominal 24 V battery. Voltage regulation eliminates transformer tap changing except between 115 V and 230 V ranges. On battery, the regulation compensates for battery voltage variations between 21.5 and 30 V. Transistorized circuits consuming little current give reasonable length of operation on battery for mobile purposes.

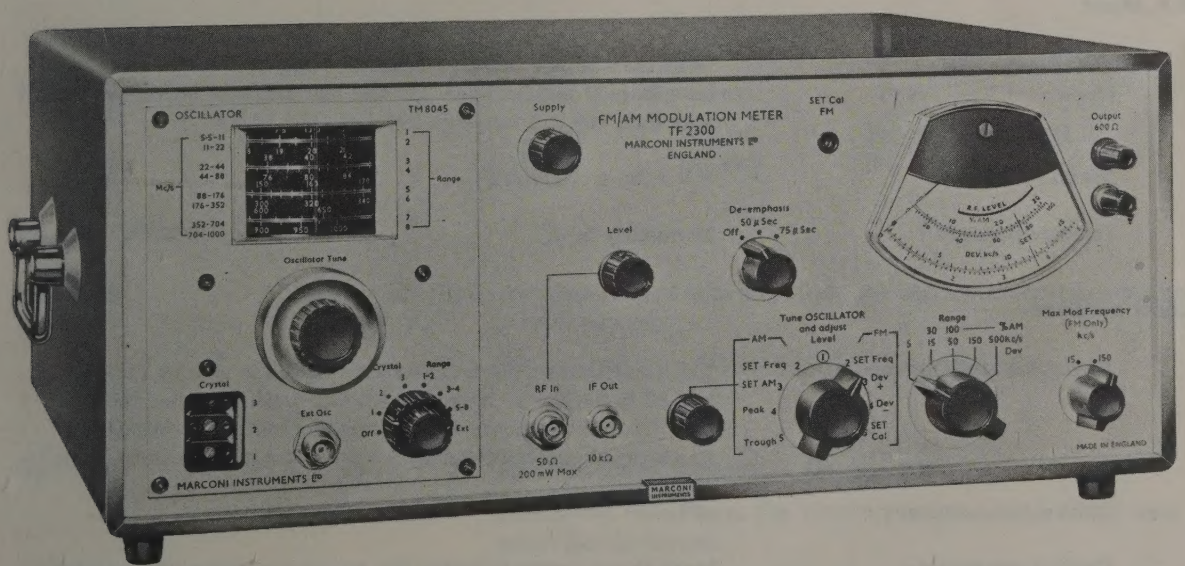


Fig. 1.1 F.M./A.M. Modulation Meter type TF 2300

1.2 DATA SUMMARY

F.M. measurement

Carrier frequency range :	4 to 1000 Mc/s.
Deviation range :	Five ranges with full-scale indications of 5, 15, 50, 150 and 500 kc/s. Positive or negative deviation indication selected by a switch.
Accuracy :	±5% of full-scale for deviations up to 500 kc/s and for modulating frequencies between 30 c/s and 150 kc/s.
A. M. rejection :	Additional deviation error less than 1 kc/s in 15 kc/s bandwidth when the a. m. depth is 80% and the modulating frequency is 1 kc/s.
Inherent noise :	Deviation less than -67 dB relative to the level of 50 kc/s deviation in a 15 kc/s bandwidth when the local oscillator is crystal controlled.

A.M. measurement

Carrier frequency range :	4 to 350 Mc/s.
Mod. depth range :	Two ranges with full-scale indications of 30% and 100% (maximum usable reading: 95%). Peak and trough indications selected by a switch.
Accuracy :	±5% of full-scale for modulating frequencies between 30 c/s and 15 kc/s.

R.F. input

Sensitivity :	Less than 20 mV between 4 and 250 Mc/s, less than 50 mV between 250 and 500 Mc/s, less than 100 mV between 500 and 1000 Mc/s.
Maximum input :	3 V r. m. s. (200 mW).
Input impedance :	Nominally 50 Ω.

Local oscillator

Variable frequency operation :	The internal oscillator covers two ranges: 5.5 to 11 Mc/s and 22 to 44 Mc/s, harmonics being used for other local oscillator frequencies. Provision is made for connecting an external oscillator with an output of approximately 20 mV.
Calibration accuracy :	±3%.
Crystal operation :	Up to three crystals within the frequency range 22 to 44 Mc/s may be fitted, for use with input frequencies between 20.5 and 1000 Mc/s.

I.F. output

Frequency :	1.5 Mc/s.
Amplitude :	Between approximately 250 and 750 mV e.m.f.
Output impedance :	Nominally 10 kΩ.

L.F. output

Frequency range :	30 c/s to 150 kc/s with switchable 15 kc/s low-pass filter on f.m. 30 c/s to 50 kc/s on a.m.
De-emphasis :	Switchable to 0, 50 or 75 μsec.
Level :	Nominally 0 dBm into 600 Ω when meter reads full-scale.
Distortion :	Less than 0.25% for f.m. deviations up to ±75 kc/s and modulation frequencies up to 15 kc/s, less than 0.5% for f.m. deviations up to ±300 kc/s and modulation frequencies up to 25 kc/s, less than 3% for f.m. deviations up to ±500 kc/s and modulation frequencies up to 150 kc/s. Less than 1% for a.m. depths up to 60%, less than 3% for a.m. depths up to 90%.

Power requirements

A.C. mains :	95 V to 130 V or 190 V to 260 V; 45 c/s to 500 c/s; 15 VA.
External battery :	21.5 V to 30 V d.c.; 320 mA at 24 V d.c.

Dimensions and weight

Height	*Width*	*Depth*	*Weight*
7 3/4 in (19 cm)	18½ in (47 cm)	14¼ in (36 cm)	24 lb (11 kg)

1.3 ACCESSORIES

Supplied

Extension Board type TM 7926; for use when servicing printed circuit boards.

Mains Lead type TM 7052.

Optional

Rack Mounting Conversion Kit type TM 8340.

Protective Cover type TM 7958/3; for protection of the front panel during transit.

Oscillator crystals. Marconi type QO1670 series; frequencies as specified in Section 2.15.

Section 2

Operation

2.1 INSTALLATION

The Modulation Meter type TF 2300 is available in two versions : with a case for bench mounting, or with a dust cover for rack mounting. The latter version, designated TF 2300R, fits a standard 19 inch rack.

Before inserting the TF 2300R into a rack, slides or runners should be fitted to the rack to support the instrument and aid location.

Transit precautions

To prevent damage being caused to the moving parts by shocks or vibrations in transit, a clip as shown in Fig. 2.1 is fitted. The tuning mechanism consists of a carriage running on a worm screw turned by the control knob; the carriage is mechanically coupled to ferrite tuning slugs in the variable inductors. The transit clip locks the carriage, and this must be released before the oscillator can be used. At the high frequency end of carriage travel, a tapped hole in the top of the brass carriage comes in line with a hole in the clip. A 6 BA screw, otherwise housed in a second hole to one side of the clip in the end plate is inserted through the clip and tightened down for long journeys, including delivery.

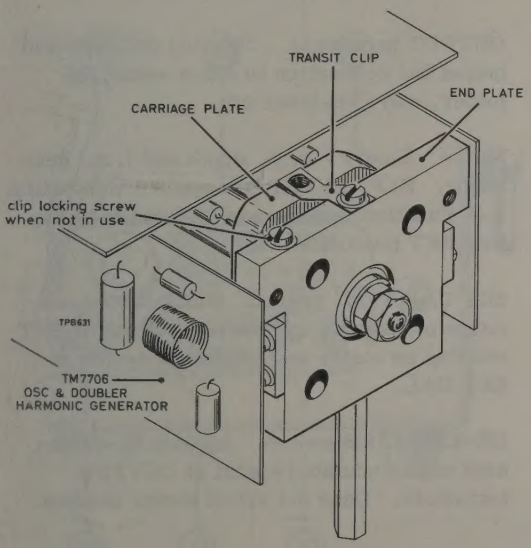


Fig. 2.1 Oscillator transit locking arrangement

To reach the clip, release the four screws at the corners of the oscillator front panel and slide the oscillator partly out. For short journeys, the carriage need not be locked if the instrument is placed upright on a layer of padding, preferably with the tuning mechanism set to the high frequency limit of its travel.

Remove the locking screw before attempting to use the oscillator.

2.2 POWER SUPPLY

The modulation meter may be used with a. c. mains or batteries.

2.2.1 Mains supply, 110–230 V

The socket of the lead provided with this instrument fits into the plug on the rear panel. The other end of the lead must be fitted with a plug by the user. The three conductors are provided with identification sleeves for this purpose, as follows :

Colour	*Connection*	*Sleeve*	*Symbol*
Blue	Line	-	-
Black	Neutral	Black	White N
Yellow/Green	Earth	Yellow	Green

Before plugging in to the mains supply, check that the fuse rating and mains voltage setting are correct. For voltages around 230 V, a 100 mA fuse is fitted; for 110 V, this must be changed to 160 mA. Check that the h. t. fuse is 500 mA and note that it is a quick-blow type. To change the mains voltage switch setting, remove the L-shaped

lock-plate secured by two screws above the switch, slide the switch button into its other position (as shown on the panel), reverse and replace the lock-plate to hold the button in its new position. No other adjustment is needed.

Switch the MAINS -BATT switch to MAINS.

2.2.2 Battery operation

The modulation meter also works on 24 V batteries. A suitable supply would be a 24 V positive earth vehicle battery, on or off charge. Negative earth systems must not be used on any account, since the case of the instrument and its coaxial connections would be live. If a positive earth supply is not available, use dry cells. Six 4.5 V batteries, e.g. Exide type H30 or 3DT9, giving 27 V initially, should give several hours of service before their output drops below 21.5 V. This is equivalent to several weeks in normal use, provided that the instrument is switched off as soon as the measurement is completed. Current drain is 350 mA approximately. Observe correct polarity when connecting batteries to the back panel terminals; left is positive.

Switch the MAINS-BATT switch to BATT.

2.3 CONTROLS

The following outline of control functions is intended for reference purposes. Until you are familiar with the instrument, use the operating procedures detailed in later sections.

① Oscillator tuning dial. Shows effective internal local oscillator frequency.

② OSCILLATOR TUNE controls. (Outer : coarse - inner : fine). Adjust for dial reading 1.5 Mc/s above signal frequency. Inoperative when using external local oscillator.

③ CRYSTAL sockets. Accept style D miniature crystals for local oscillator control.

④ EXT OSC socket. Type BNC. Accepts external local oscillator signal of about 100 mV.

⑤ CRYSTAL switch. Selects crystal required. Switch to OFF when not crystal controlled.

⑥ RANGE switch. Selects internal local oscillator ranges as shown beside tuning dial, or external oscillator.

⑦ R.F. IN socket. Type BNC. Accepts signal under test.

⑧ I.F. OUT socket. Type BNC. Connect to counter for measuring carrier shift or to oscilloscope for viewing unfiltered a.m. envelope. Output is 1.5 Mc/s at about $\frac{1}{2}$ V.

⑨ LEVEL control. Adjusts attenuation of signal input at R.F. IN socket.

⑩ SET A.M. level control. Adjust for SET reading on meter when Function selector is at SET A.M.

⑪ Function switch. Selects setting up and measuring conditions on f.m. and a.m. Numbers show normal sequence of settings.

⑫ A.M./DEV RANGE switch. Selects full-scale meter range, five for f.m., two for a.m.

⑬ MAX MOD FREQ switch. Selects modulation bandwidth in f.m. measurements.

⑭ OUTPUT terminals. Provide demodulated output for connection to extra-sensitive meter, wave analyser etc.

⑮ Meter. Reads % a.m. depth and f.m. deviation. R.F. LEVEL arc applies to position 1 of Function switch, and SET mark to the four SET positions.

⑯ SET CAL F.M. preset. Standardizes deviation measuring accuracy. Adjust for SET reading on meter with Function switch at SET CAL.

⑰ DE-EMPHASIS switch. Applies de-emphasis to demodulated signal at OUTPUT terminals. Does not affect meter readout.

⑱ SUPPLY switch. Shows red segments when on. Works on both mains and battery operation.

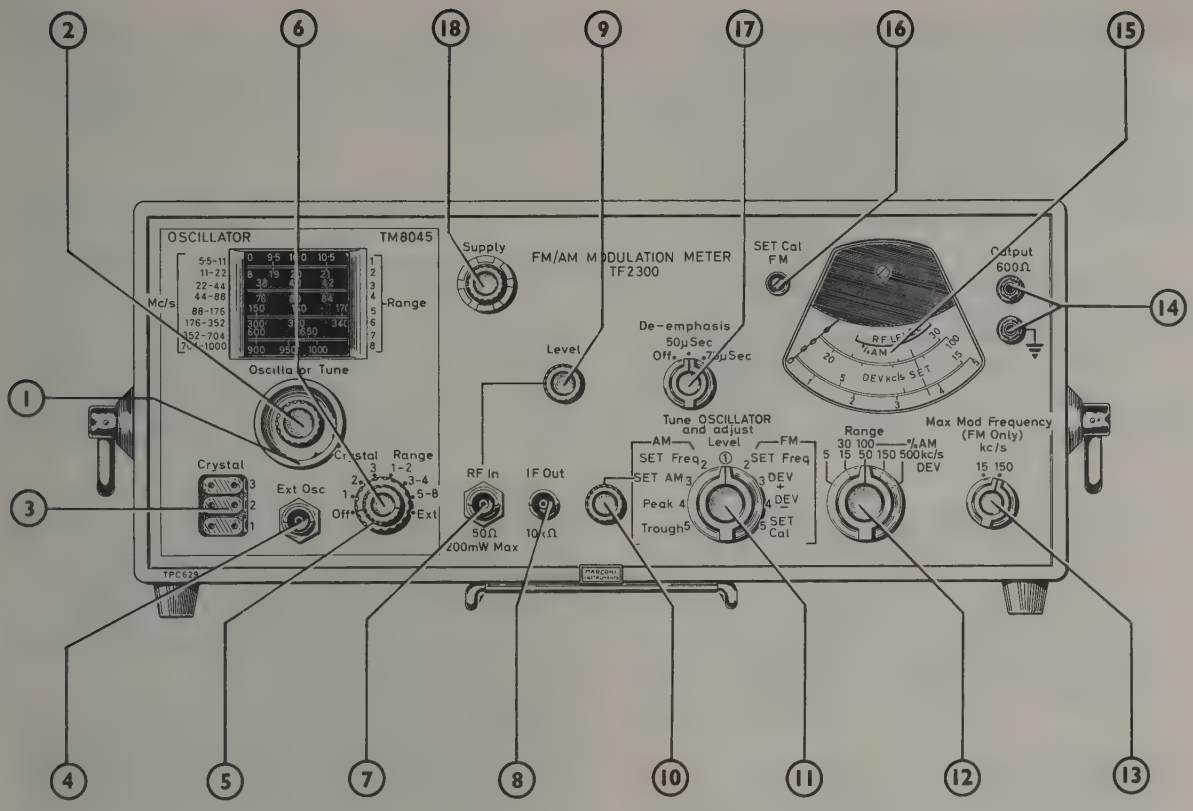


Fig. 2.2 Front panel controls

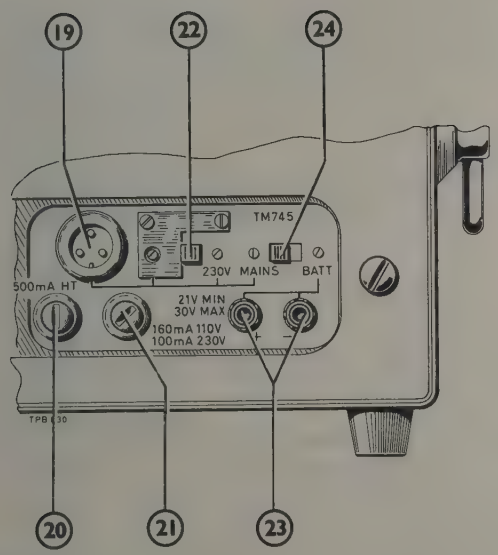


Fig. 2.3 Rear panel

⑲ Supply plug. Accepts Mains Lead type TM 7052 for a.c. operation.

⑳ H.T. fuse. 500 mA quick-blow type.

㉑ Mains fuse. 100 mA (for 230 V a.c.) or 160 mA (for 110 V a.c.) slow-blow type.

㉒ Mains voltage selector. 230 V position for 190-260 V, 110 V position for 90-130 V. Reverse locking plate when changing range.

㉓ Battery terminals. 21-30 V d.c. floating or positive earth.

㉔ MAINS/BATT switch. Select MAINS for a.c. supply to plug, BATT for d.c. supply to terminals.

2.4 PREPARATION FOR USE

It is helpful in reading these instructions and in using the instrument until completely familiar with its operation, to remember that where a control or switch position is marked SET, the control should be adjusted to make the meter read on the line marked SET.

Before commencing measurements :

(1) Check the mechanical zero of the meter and adjust, if necessary, by means of the screw.

(2) Turn the LEVEL control towards minimum (counter-clockwise).

(3) Connect the signal to be tested to the R. F. IN socket at a level which, operating into 50 Ω, will develop a voltage within the limits given in section 1.2 under 'R. F. input'. At the highest carrier frequencies the shortest possible input lead should be used or an external 50 Ω matching attenuator.

CAUTION Inputs greater than 3 V r.m.s. may damage the attenuator. If in doubt, use an external attenuator.

(4) Turn the SUPPLY switch clockwise so that the red segments show.

2.5 MEASURING F.M. DEVIATION

After carrying out the preparation procedure in section 2.4 :

(1) Set the A. M./DEV RANGE switch (below the meter) to suit the expected deviation of the signal under test.

(2) Set the MAX MOD FREQUENCY switch to 150 kc/s unless the modulating frequency is less than 15 kc/s. If so, switch to 15 kc/s in order to obtain the best signal-to-noise ratio.

(3) Set the DE-EMPHASIS switch as required. Its setting has no effect on the internal meter reading but it is effective on the demodulated signal at the OUTPUT terminals.

TABLE 2.1

Frequency ranges

RANGE switch setting	Range	Osc. freq. Mc/s	R.F. input freq. Mc/s
1-2	1	5.5 - 11	4.0 - 9.5
	2	11 - 22	9.5 - 20.5
3-4	3	22 - 44	20.5 - 42.5
	4	44 - 88	42.5 - 86.5
5-8	5	88 - 176	86.5 - 174.5
	6	176 - 352	274.5 - 350.5
	7	352 - 704	350.5 - 702.5
	8	701.5 - 1001.5	700 - 1000
EXT	-	F + 1.5	F

(4) Set the oscillator RANGE switch in accordance with table 2.1. For detailed instructions on oscillator arrangements, see section 2.10. In this section, it is assumed that the internal local oscillator will be used.

(5) Set the Function switch to TUNE OSCILLATOR AND ADJUST LEVEL. Adjust the OSCILLATOR TUNE dial to a frequency 1.5 Mc/s above the carrier. Tune for peak meter deflection, and then adjust LEVEL for a meter deflection in the black R. F. LEVEL arc, preferably near the top end for best signal-to-noise ratio.

(6) Set the Function switch to F. M. SET FREQ. Slightly readjust the OSCILLATOR TUNE control until the meter deflects to the SET line. Avoid spurious settings - the correct one is the closest to the original setting obtained in (5).

(7) Set the Function switch to DEV + and DEV-. The demodulated peak frequency deviations above and below the carrier frequency may be measured on the meter or viewed on an oscilloscope connected to the OUTPUT terminals. Read the meter on the scale corresponding to the setting of the A. M./DEV RANGE switch.

Note : If unequal readings are obtained, distortion in the equipment under test is indicated. See section 2.12.

(8) Before carrying out a series of measurements, and from time to time during measurements,

switch to SET CAL and check that the meter reads SET. If necessary, adjust the f.m. calibration by inserting a screwdriver in the SET CAL F.M. preset and adjusting until the meter reads SET.

2.6 MEASURING A.M. DEPTH

After carrying out the preparation procedure in section 2.4 :

(1) Set the A.M./DEV RANGE switch (below the meter) to the 30% or 100% a.m. depth range according to the expected modulation depth of the signal under test.

(2) Set the oscillator RANGE switch in accordance with table 2.1. For detailed instructions on oscillator arrangements, see section 2.10. In this section, it is assumed that the internal local oscillator will be used.

(3) Set the Function switch to TUNE OSCILLATOR AND ADJUST LEVEL. Adjust the OSCILLATOR TUNE dial to a frequency of 1.5 Mc/s above the input frequency. Tune for peak deflection and then adjust LEVEL to give a meter deflection in the black R.F. LEVEL arc, preferably towards the lower end for best mixing and hence minimum distortion.

(4) Set the Function switch to A.M. - SET FREQ. Slightly readjust the OSCILLATOR TUNE control until the meter deflects to the SET line. Avoid spurious settings - the correct one is the closest to the original setting obtained in (3).

(5) Set the Function switch to SET A.M. Adjust the adjacent SET A.M. control until the meter reads on the SET line. Note that this setting can be done with or without modulation since the adjustment is to the mean level of the i.f. signal. However, appreciable even harmonic distortion can affect the setting accuracy and, in such circumstances, it is preferable to set up without modulation.

(6) Set the Function switch to A.M. PEAK. Read the percentage modulation depth at the peak.

(7) Set the Function switch to A.M. TROUGH. Read the percentage modulation depth at the trough.

When the peak and trough readings are unequal, distortion is present in the input signal.

2.7 MEASURING F.M. ON A.M.

To measure the spurious f.m. on an amplitude modulated source, proceed as for f.m. measurement, section 2.5.

A certain amount of spurious f.m. is introduced by the instrument iself. When the carrier is deeply modulated, at high modulating frequencies, this should be taken into consideration, but below 40% depth in the audio range of frequencies the spurious deviation is not very significant. At 80% depth and 1 kc/s frequency, the maximum spurious deviation is less than 500 c/s referred to the 15 kc/s bandwidth.

Tables 2.2 and 2.3 give typical results which should be used for general guidance purposes only .

TABLE 2.2

Spurious deviations with increasing % a.m.

1. LEVEL control set for meter reading at top end of arc.
2. A.M. frequency = 1 kc/s

	Spurious deviation	
A.M.	15 kc/s bandwidth	150 kc/s bandwidth
30	< 500 c/s	< 1 kc/s
80	< 500 c/s	2 kc/s

TABLE 2.3

Spurious deviations with increasing a.m. frequency

1. LEVEL control set for meter reading at top end of arc.
2. A.M. depth = 80%.

	Spurious deviation	
A.M. freq.	15 kc/s bandwidth	150 kc/s bandwidth
1 kc/s	< 500 c/s	2 kc/s
10 kc/s	1 kc/s	3 kc/s
100 kc/s	-	7 kc/s

2.8 MEASURING A.M. ON F.M.

The modulation meter may be used to indicate amplitude modulation in a frequency modulated signal, provided that the deviation is less than 100 kc/s. The procedure is that of a. m. measurement, section 2. 6.

In general, the spurious a. m. indication will be proportional to the deviation. The i. f. amplifier is set up for optimum phase response for f. m. deviation measurement rather than a maximally flat response for measuring a. m. on f. m.

Below 100 kc/s deviation, the internally generated spurious a. m. is less than 2% approximately.

2.9 NOISE MEASUREMENTS

By connecting an external meter to the OUTPUT terminals, noise measurements limited only by the noise level generated within the instrument can be made. The internal meter, being peak reading, is not suitable for measuring noise and, ideally, an r. m. s. responding meter should be used. However, sufficiently accurate results are usually obtained with an average reading meter by applying the appropriate correction factor.

Typical meters which can be used are as follows :

(a) R. M. S. valve voltmeter capable of measuring to the necessary accuracy (1% deviation on any range is approximately 7. 75 mV). Errors due to crest factor and zero shift can be avoided by operating the meter at mid-scale by means of an attenuator.

(b) Marconi Instruments Sensitive Valve Voltmeter, type TF 2600.

(c) Marconi Instruments Distortion Factor Meter, type TF 2331.

(b) and (c) are average reading and a correction of +1 dB should be applied.

(1) The modulation meter has two selectable f. m. bandwidths and the appropriate filter must be selected when making the test. For optimum results, a capacitor should be connected across the OUTPUT terminals - 0. 014 μF for the 15 kc/s bandwidth, or 0. 0013 μF for 150 kc/s. These corrective capacitors are necessary because of the design of the filters (see section 3. 9).

(2) The output of the modulation meter is 0. 775 V into 600 Ω for full-scale deflection on the internal meter.

(3) For measurements in the r. f. range 22 Mc/s to 1000 Mc/s the local oscillator must be crystal controlled if lowest possible internally generated noise is required. Thus, a crystal suitable for the appropriate frequency must be available.

This is not a requirement below 22 Mc/s, where the internal noise of the oscillator is sufficiently low to make crystal control unnecessary.

(4) The curves given in Fig. 2. 4 show typical noise levels produced by free-running and crystal controlled oscillators over the r. f. range of the instrument. In the 15 kc/s bandwidth, below 70 to 100 Mc/s, there is insignificant difference in respective noise levels. Above 100 Mc/s, however, the noise level free-running increases progressively with frequency, whereas under crystal control the level remains nearly constant. The necessity to use crystal control depends on the noise level of the equipment under test and, as shown by the curves, the bandwidth in which the measurement is to be made and the carrier frequency.

In the 150 kc/s bandwidth, there is little advantage in using crystal control below 500 Mc/s. In both bandwidths, crystal control will eliminate oscillator microphony and therefore may be advantageous in conditions of vibration or high acoustic noise levels.

F.M. noise

(1) Connect an external meter to the OUTPUT terminals. Apply r. f. input at a suitable level.

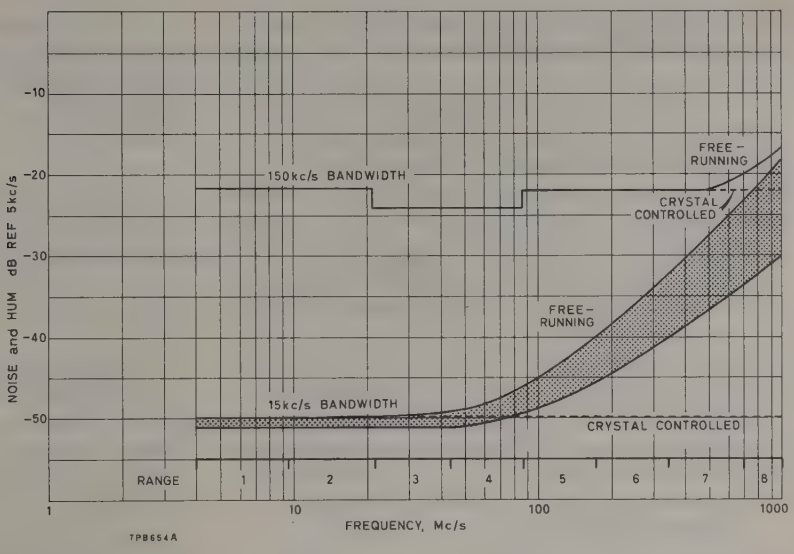


Fig. 2.4 Modulation meter noise levels

(2) Adjust the OSCILLATOR TUNE dial to a frequency 1.5 Mc/s above the carrier and tune for peaking, as in section 2.5 (5). Then adjust LEVEL to the top end of the black arc on internal meter. Switch to 15 kc/s or 150 kc/s MAX MOD FREQUENCY setting, thus selecting the required low-pass filter in the modulation meter.

(3) Switch to position F.M. SET FREQ and adjust the oscillator until meter reads SET. Switch to DEV +.

(4) Switch to crystal control and check that the meter still reads near to the SET mark - the actual reading will depend on frequency tolerances in the source and local oscillator output.

(5) Modulate the carrier to the reference deviation and select the appropriate range on the A.M./DEV RANGE switch.

(6) Read the external meter and record the dB level.

(7) Remove the modulation from the signal under test. Adjust the meter sensitivity, obtain and record the new level in dB.

(8) The signal-to-noise ratio referred to the selected bandwidth in which the measurements were made is given by the difference between the levels measured in (6) and (7).

A.M. noise

(1) Connect an external meter to the OUTPUT terminals. Apply r.f. input at a suitable level.

(2) Adjust the OSCILLATOR TUNE dial to a frequency 1.5 Mc/s above the carrier and tune for peaking, as in section 2.6 (3). Then adjust LEVEL to the top end of the black arc on internal meter.

(3) Switch to position A.M. SET FREQ and adjust the oscillator until the meter reads SET. Switch to A.M. PEAK.

(4) Switch to crystal control and check that the meter still reads near to the SET mark - the actual reading will depend on frequency tolerances in the source and local oscillator output.

(5) Modulate the carrier to the reference depth and select the appropriate range on the A.M./DEV RANGE switch.

(6) Read the external meter and record the dB level.

(7) Remove the modulation from the signal under test. Adjust the meter sensitivity, obtain and record the new level in dB.

(8) The signal-to-noise ratio is given by the difference between the levels measured in (6) and (7).

2.10 OSCILLATOR ARRANGEMENTS AND OPERATING NOTES

The same procedure applies for both f. m. and a. m. measurements. The basic requirement is for a stable frequency source 1.5 Mc/s higher than the signal under test, and this can be obtained from three sources :

(a) The internal permeability-tuned variable oscillator usable at all frequencies between 4 and 1000 Mc/s.

(b) The internal crystal controlled spot-frequency oscillator, usable with selected crystals over the range of 20 to 1000 Mc/s.

(c) An external oscillator with an output of 100 mV across 50 Ω.

2.10.1 Variable oscillator

Choose the appropriate range on the RANGE switch.

Check that the CRYSTAL selector is at OFF.

Rotate the TUNE dial with the coarse and fine controls to a setting 1.5 Mc/s above the frequency of the input signal. Note that there are meter peaks 1.5 Mc/s above and below the carrier frequency; always tune the oscillator to the higher frequency for correct polarity sense of the Function switch.

Adjust the fine control from time to time, to bring the meter reading to the SET mark when switched to the SET FREQ position. It is permissible to tune to another internal oscillator harmonic as this will not affect the readings obtained.

The oscillator takes about two minutes to stabilize after turning on, but this is only of any significance at very high frequencies.

2.10.2 Crystal-controlled oscillator

Plug in a crystal of the required frequency as selected in section 2.15.

Select the appropriate position on the RANGE switch and set the oscillator dial to the crystal frequency. Set the CRYSTAL selector to the crystal position number.

Slight tuning on position TUNE OSCILLATOR of the Function switch may be necessary to obtain maximum sensitivity at the higher frequencies. Do not tune far off the original setting to avoid any possibility of mode jumping.

Proof that the oscillator is under crystal control can be obtained by setting the Function switch to SET FREQ and rocking the tuning control. If the oscillator is under control the meter reading will not vary.

In general, the tuning control is set to the required oscillator frequency, but provided the crystal frequencies do not differ too widely, any of the crystals can be selected without re-setting the tuning control between selections.

2.10.3 External oscillator

If a stable frequency source is available, it may be preferable to use it rather than the internal oscillator. Feed the external oscillator signal into the EXT OSC socket and turn the RANGE switch to EXT. An input level of 100 mV into 50 Ω is required at a frequency 1.5 Mc/s above the input signal frequency.

The primary use of this facility is for measurement on r.f. signals which contain harmonics of less than 20 dB down on the fundamental. Because the internal oscillator signal also contains harmonics, mixing two such signals can give rise to spurious results. A 'pure' local oscillator signal prevents this happening.

Secondly, if an external crystal oscillator is available, it may be preferred to use this, rather than to crystal control the local oscillator, for low level noise measurements.

Thirdly, the carrier frequency range may be extended from 1000 Mc/s to approximately 2500 Mc/s by using a suitable external oscillator. In this range, however, the sensitivity may deteriorate.

2.11 MEASUREMENT IN 1 TO 2 Mc/s RANGE

The instrument can accept signals in the band 1 to 2 Mc/s, applied directly to R. F. IN, with the oscillator switched off. Input sensitivity is the same as usual. Ensure the modulation

frequency does not take the frequency beyond the band limits quoted, so as to avoid introducing distortion caused by the reduced gain outside the pass band.

Operating the instrument in this way cuts out the frequency inversion caused by the heterodyne system normally used. The polarity of the DEV + and DEV - positions of the Function switch is therefore reversed.

2.12 ASYMMETRIC MODULATION AND CARRIER SHIFT

The presence of asymmetric modulation usually indicates distortion in the equipment under test, and is revealed by unequal meter readings when the Function switch is turned between PEAK and TROUGH or DEV+ and DEV-.

To make sure that the asymmetry is due to the input signal, and not introduced by the instrument, retune the oscillator to the lower peak, 1.5 Mc/s below the carrier frequency. If the asymmetry is reversed, i.e., the original DEV + reading now appears at DEV -, and vice versa, the asymmetric modulation can be attributed to the input signal.

In position SET FREQ, the output from a counter circuit in the limiter is measured by the meter. When the meter indicates SET, the i.f. is centred on 1.5 Mc/s exactly. If modulation then causes the meter reading to fluctuate, carrier shift is present. The extent of the shift may be measured by connecting a counter to I.F. OUT. In cases of severe carrier shift, the modulation meter must be set up with modulation on.

2.13 USE OF L.F. OUTPUT TERMINALS

CAUTION Application of d.c. to these terminals will cause damage - see section 3.10.

These terminals enable the modulation waveform on the input signal to be monitored or analysed. They also enable a sensitive external meter to be used to extend the deviation range down to the level of residual noise within the instrument.

The terminals are fed by an independent output stage in the 2nd l.f. amplifier and therefore the internal meter is unaffected by loads connected to them.

Depending on the position of the Function switch, a.m. or f.m. demodulated outputs are obtained. The output impedance is approximately 600 Ω and the open circuit level approximately 1.5 V for full-scale deflection on the internal meter.

Note : For most purposes, the output can be terminated with 600 Ω or unterminated. In the latter case, a voltage due to leakage of C11 may appear at the output, causing difficulty if a d.c. coupled oscilloscope is used at high sensitivity.

The l.f. response is substantially level up to 150 kc/s. The output is also available with switched 50 μsec or 75 μsec de-emphasis to restore the modulation characteristic of signals that have had pre-emphasis applied.

2.14 USE OF I.F. OUT SOCKET

This socket can be used to measure carrier shift by connecting a counter to it - see section 2.6 - or to view the a.m. envelope on an oscilloscope.

It may also be used to measure amplitude modulation above 50 kc/s which normally would not be passed by the a.m. detector. This is an extended use of the modulation meter which in practice would probably be limited by the attenuation of the 10 kΩ output resistance and the input capacitance of the measuring instrument.

2.15 CRYSTAL SELECTION

Crystal control of the local oscillator reduces microphony and pick-up from surrounding equipment where these effects cannot be avoided. At higher frequencies, distortion due to noise in the oscillator can be significantly reduced. Up to three crystals can be plugged into sockets on the oscillator front panel and any one can be selected by the red coloured section of the oscillator RANGE switch.

A recommended crystal is a 3rd overtone series resonant Marconi Type QO1670F, with a specified frequency between 22 to 44 Mc/s. Thus, the crystal is used on its fundamental on Range 3 and appropriate harmonics on the other and higher ranges. These crystals are in hermetically sealed, style D miniature cases conforming to British Standard and U.S. Style HC6U.

The following simplifies the procedure for establishing the required crystal frequency for any specified r.f. input between 20.5 Mc/s and 1000 Mc/s :

1) Add 1.5 Mc/s to the r.f. input frequency.

2) Divide this sum by the local oscillator harmonic factor found from Table 2.4 (shown opposite the range applicable to the r.f. input). This result is the crystal frequency.

3) Expressing 1) and 2) as a formula :

$$\text{Crystal frequency} = \frac{(\text{R.F.} + 1.5)\ \text{Mc/s}}{\text{Harmonic factor}}$$

TABLE 2.4

Range no.	*R.f. input frequency Mc/s*	*Local oscillator harmonic factor*
1	4 - 9.5	No crystal control on this range
2	9.5 - 20.5	No crystal control on this range
3	20.5 - 42.5	1
4	42.5 - 86.5	2
5	86.5 - 174.5	4
6	174.5 - 350.5	8
7	350.5 - 702.5	16
8	702.5 - 1000	24

Example : To find the fundamental crystal frequency for an r.f. input of 83.666 Mc/s :

1) Consulting Table 2.4, it can be seen that this r.f. falls within range 4 and the harmonic factor is therefore 2.

$$\text{Crystal frequency} = \frac{83.666 + 1.5}{2} = 42.583\ \text{Mc/s}$$

A ±10% tolerance is permitted on the 1.5 Mc/s i.f. (±150 kc/s) and, since the normal tolerance on crystal frequency for +20°C to +60°C temperature operation is ±0.01%, a crystal with a frequency of 42.58 Mc/s and normal tolerance should be suitable and it is not usually necessary to specify a tighter tolerance.

Note : Where f.m. deviation approaches the full i.f. bandwidth and the r.f. approaches 1000 Mc/s, the crystal tolerance should be as close as possible for minimum distortion and crystals with tighter frequency limits (±0.003%) should be specified.

In every case the crystal frequency selected should be such that the lowest possible harmonic factor gives the required operating frequency. This is automatically given by Table 2.4 using the procedure described.

Example : Required oscillator frequency is 176 Mc/s.

1) It can be seen that a crystal of 22 Mc/s operating with a harmonic factor of 8 will satisfy frequency conditions.

2) However, a crystal of 44 Mc/s operating with a harmonic factor of 4 should be specified for maximum output from the oscillator.

Crystals are fully specified in Data Sheet QO1670F, to which reference should be made for further details if required. Both data sheet and crystals are available from The Marconi Co. Ltd., Chelmsford, Essex.

2.16 F.M. STEREO MEASUREMENTS

Distortion and channel separation in stereo systems which are designed to meet U.S. Federal Communications Commission (F.C.C.) requirements can be measured with the TF 2300 Modulation Meter. The general procedures are described in the following sections.

2.16.1 Harmonic distortion

It can be assumed that distortion introduced by the modulation meter is small in comparison with that in stereo transmitters, being typically

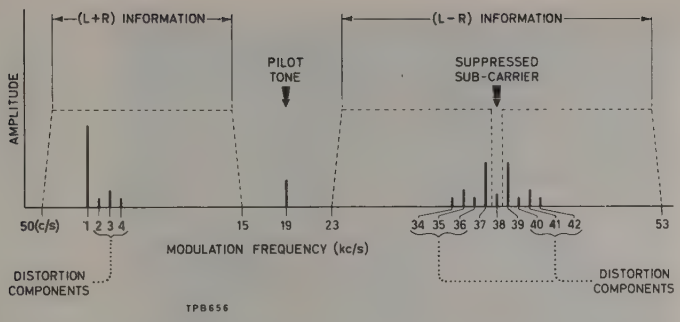


Fig. 2.5 Frequency spectrum of f.m. stereo signal with 1 kc/s modulating tone

better than 0.2% for deviations up to ±75 kc/s and therefore, for practical purposes, can be ignored.

The typical frequency spectrum of a G.E.-Zenith type of f.m. stereo signal when a 1 kc/s test tone (for example) is applied to the left channel and no signal to the right, is shown in Fig. 2.5.

Ideally, only the fundamental 1 kc/s signal in the left + right (L + R) channel and the 37 and 39 kc/s sidebands of the suppressed sub-carrier in the L-R channel should be present. In practice, however, harmonic distortions will occur at 2, 3, 4 kc/s .... in the L + R channel and at 37 and 39, 36 and 40, 35 and 41 kc/s .... in the L - R channel. These are the distortion components which it is desired to measure.

The test arrangement for measuring distortion in an f.m. stereo transmitter is shown for measurements on the G.E.-Zenith system is given below :-

(1) Connect the transmitter output to the modulation meter in accordance with section 2.4 - Preparation for Use.

(2) Set up the modulation meter with the transmitter signal on, in accordance with section 2.5 - Measuring F.M. Deviation.

The following notes are intended to amplify the setting instructions and refer specifically to the G.E.-Zenith System.

(3) Set the controls as follows :-

A.M./DEV RANGE : 150 kc/s.

MAX MOD FREQ : 150 kc/s.

DE-EMPHASIS : OFF, even if pre-emphasis is being applied.

Function : DEV + or DEV -.

(4) Apply a test signal of a chosen frequency to the left channel with no signal to the right channel.

(5) Connect Wave Analyser type TF 2330, or an equivalent instrument for harmonic analysis of the audio range, to the OUTPUT terminals of the modulation meter. (Refer to the instruction manual on the analyser for calibration and measuring procedures.)

(6) The frequency spectrum of the demodulated output of the modulation meter can be obtained by tuning through the frequency range and taking measurements at the appropriate frequencies. The result is usually referred to the fundamental and expressed as percentages or dB readings, calculated by the usual method for harmonic distortion measurement as described in the wave analyser manual.

For practical working purposes, the small amount of distortion introduced by the modulation meter can be neglected.



Fig. 2.6 Test arrangement for measuring distortion

2.16.2 Channel separation

If it were possible to view the 50 c/s to 15 kc/s and the 23 to 53 kc/s regions of Fig. 2.5 separately, the waveforms shown in Figs. 2.7 (a) and (b) would be seen.

The ideal composite modulating signal with the 19 kc/s pilot tone removed should be the sum of these waveforms as shown in (c). In practice, there will be amplitude unbalance and relative phase shift (group delay error) between the two added signals. These two effects, which can be viewed on a direct coupled oscilloscope, are illustrated in (d) and (e), where amplitude 'B' represents a signal on the right channel due to an input on the left channel.

Channel 1 separation is given by :

$$\text{Separation} = 20 \log_{10} \frac{A}{B} \quad \ldots\ldots (1)$$

where A and B are measurements indicated in Figs. 2.7 (d) and (e).

An f.m. stereo transmitter or stereo modulator must meet specified requirements as regards response and phase shift. In the F.C.C. requirements, it is deemed that these are met if the channel separation exceeds 29.7 dB.

A suitable oscilloscope, such as a Marconi Instruments type TF 2200, connected to the OUTPUT terminals of the modulation meter, enables the channel separation to be measured. Fig. 2.8 shows typical oscillograms obtained on a TF 2200 Oscilloscope from which such measurements would be made.

From 200 c/s to 15 kc/s, the channel separation figure of the modulation meter is typically 13 dB below the minimum required of f.m. stereo channels in the F.C.C. regulations, while at 50 c/s (the worst condition), it is at least 3 dB below.

The test arrangement for measuring channel separation is as illustrated in Fig. 2.6, except that the wave analyser is now replaced by an oscilloscope, and the procedure is as follows :-

(1) Set up the equipment and modulation meter as already described for distortion measurements.

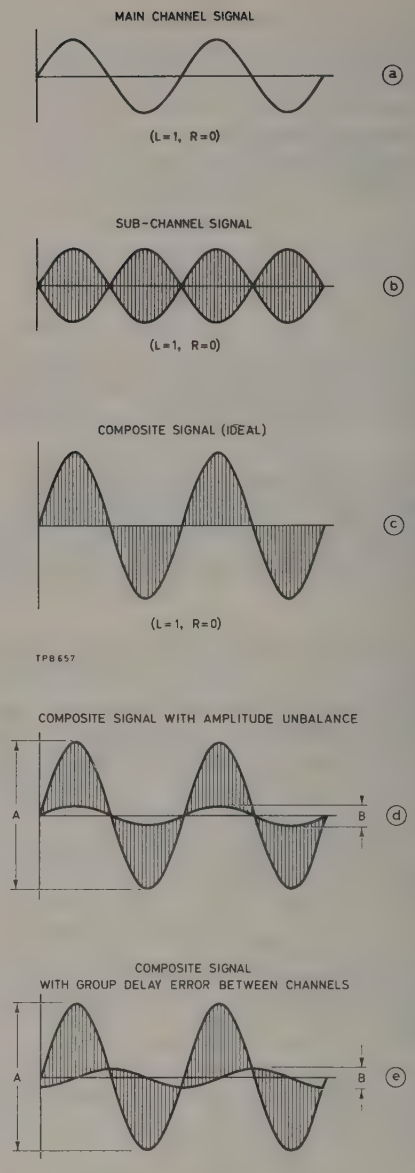


Fig. 2.7 Derivation of composite stereo signal showing unbalance and phase shift errors

(2) Apply a test signal to the left channel at 15 kc/s modulation frequency (for example) with no signal on the right channel.

(3) Connect the oscilloscope, switched to d.c. coupling, to the OUTPUT terminals of the modulation meter. If the oscilloscope has a high input resistance it is advisable to connect a resistor of about 1 kΩ (the exact

value is unimportant) in parallel with the oscilloscope input; this is to prevent leakage in the electrolytic output coupling capacitor of the modulation meter from producing a voltage which may shift the trace off the screen. Adjust the oscilloscope to suitable time base and input attenuator settings for a composite signal waveform as in Fig. 2.8. Procedure here depends on the equipment being tested, in general it consists of setting up the system for deviation, and gain etc., in accordance with individual system manufacturers' instructions until a satisfactory waveform is seen. The gain of the L - R channel should then be adjusted for minimum amplitude 'B'. Measure this amplitude and calculate separation in accordance with formula (1).

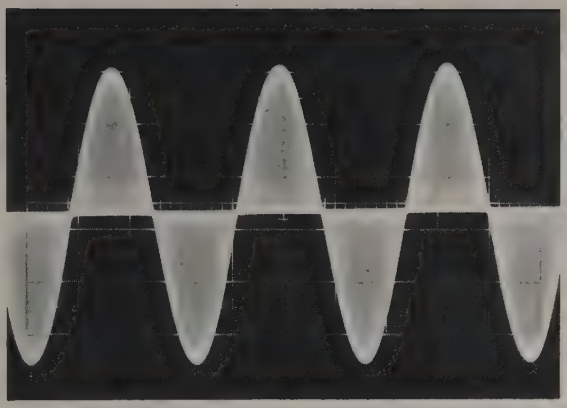


Fig. 2.8 (a) Ideal

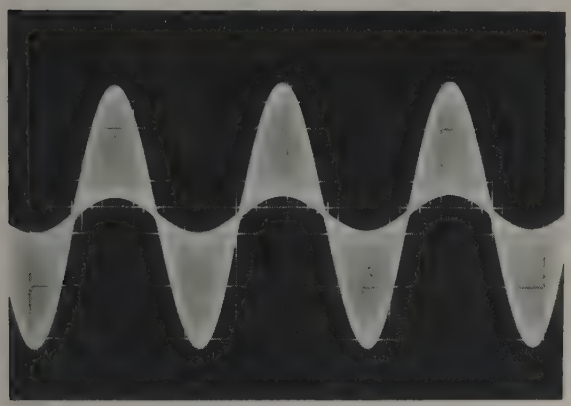


Fig. 2.8 (b) With amplitude unbalance

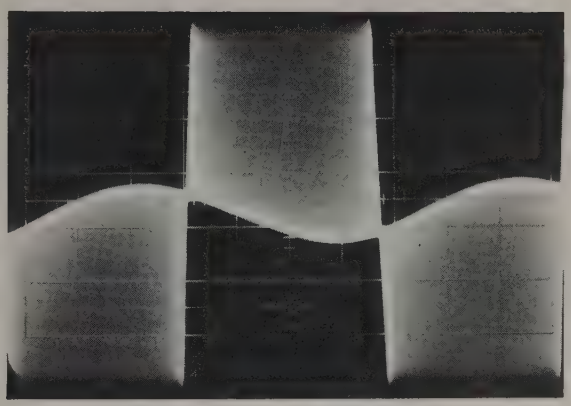


Fig. 2.8 (c) With phase shift error

Oscillograms of composite signal

DECIBEL CONVERSION TABLE

Ratio Down			Ratio Up	
VOLTAGE	POWER	DECIBELS	VOLTAGE	POWER
1·0	1·0	**0**	1·0	1·0
·9886	·9772	**·1**	1·012	1·023
·9772	·9550	**·2**	1·023	1·047
·9661	·9333	**·3**	1·035	1·072
·9550	·9120	**·4**	1·047	1·096
·9441	·8913	**·5**	1·059	1·122
·9333	·8710	**·6**	1·072	1·148
·9226	·8511	**·7**	1·084	1·175
·9120	·8318	**·8**	1·096	1·202
·9016	·8128	**·9**	1·109	1·230
·8913	·7943	**1·0**	1·122	1·259
·8710	·7586	**1·2**	1·148	1·318
·8511	·7244	**1·4**	1·175	1·380
·8318	·6918	**1·6**	1·202	1·445
·8128	·6607	**1·8**	1·230	1·514
·7943	·6310	**2·0**	1·259	1·585
·7762	·6026	**2·2**	1·288	1·660
·7586	·5754	**2·4**	1·318	1·738
·7413	·5495	**2·6**	1·349	1·820
·7244	·5248	**2·8**	1·380	1·905
·7079	·5012	**3·0**	1·413	1·995
·6683	·4467	**3·5**	1·496	2 239
·6310	·3981	**4·0**	1·585	2·512
·5957	·3548	**4·5**	1·679	2·818
·5623	·3162	**5·0**	1·778	3·162
·5309	·2818	**5·5**	1·884	3·548
·5012	·2512	**6**	1·995	3·981
·4467	·1995	**7**	2·239	5·012
·3981	·1585	**8**	2·512	6·310
·3548	·1259	**9**	2·818	7·943
·3162	·1000	**10**	3·162	10·000
·2818	·07943	**11**	3·548	12·59
·2512	·06310	**12**	3·981	15·85
·2239	·05012	**13**	4·467	19·95
·1995	·03981	**14**	5·012	25·12
·1778	·03162	**15**	5·623	31·62

DECIBEL CONVERSION TABLE (continued)

Ratio Down			Ratio Up	
VOLTAGE	POWER	DECIBELS	VOLTAGE	POWER
·1585	·02512	**16**	6·310	39·81
·1413	·01995	**17**	7·079	50·12
·1259	·01585	**18**	7·943	63·10
·1122	·01259	**19**	8·913	79·43
·1000	·01000	**20**	10·000	100·00
·07943	$6{\cdot}310 \times 10^{-3}$	**22**	12·59	158·5
·06310	$3{\cdot}981 \times 10^{-3}$	**24**	15·85	251·2
·05012	$2{\cdot}512 \times 10^{-3}$	**26**	19·95	398·1
·03981	$1{\cdot}585 \times 10^{-3}$	**28**	25·12	631·0
·03162	$1{\cdot}000 \times 10^{-3}$	**30**	31·62	1,000
·02512	$6{\cdot}310 \times 10^{-4}$	**32**	39·81	$1{\cdot}585 \times 10^{3}$
·01995	$3{\cdot}981 \times 10^{-4}$	**34**	50·12	$2{\cdot}512 \times 10^{3}$
·01585	$2{\cdot}512 \times 10^{-4}$	**36**	63·10	$3{\cdot}981 \times 10^{3}$
·01259	$1{\cdot}585 \times 10^{-4}$	**38**	79·43	$6{\cdot}310 \times 10^{3}$
·01000	$1{\cdot}000 \times 10^{-4}$	**40**	100·00	$1{\cdot}000 \times 10^{4}$
$7{\cdot}943 \times 10^{-3}$	$6{\cdot}310 \times 10^{-5}$	**42**	125·9	$1{\cdot}585 \times 10^{4}$
$6{\cdot}310 \times 10^{-3}$	$3{\cdot}981 \times 10^{-5}$	**44**	158·5	$2{\cdot}512 \times 10^{4}$
$5{\cdot}012 \times 10^{-3}$	$2{\cdot}512 \times 10^{-5}$	**46**	199·5	$3{\cdot}981 \times 10^{4}$
$3{\cdot}981 \times 10^{-3}$	$1{\cdot}585 \times 10^{-5}$	**48**	251·2	$6{\cdot}310 \times 10^{4}$
$3{\cdot}162 \times 10^{-3}$	$1{\cdot}000 \times 10^{-5}$	**50**	316·2	$1{\cdot}000 \times 10^{5}$
$2{\cdot}512 \times 10^{-3}$	$6{\cdot}310 \times 10^{-6}$	**52**	398·1	$1{\cdot}585 \times 10^{5}$
$1{\cdot}995 \times 10^{-3}$	$3{\cdot}981 \times 10^{-6}$	**54**	501·2	$2{\cdot}512 \times 10^{5}$
$1{\cdot}585 \times 10^{-3}$	$2{\cdot}512 \times 10^{-6}$	**56**	631·0	$3{\cdot}981 \times 10^{5}$
$1{\cdot}259 \times 10^{-3}$	$1{\cdot}585 \times 10^{-6}$	**58**	794·3	$6{\cdot}310 \times 10^{5}$
$1{\cdot}000 \times 10^{-3}$	$1{\cdot}000 \times 10^{-6}$	**60**	1,000	$1{\cdot}000 \times 10^{6}$
$5{\cdot}623 \times 10^{-4}$	$3{\cdot}162 \times 10^{-7}$	**65**	$1{\cdot}778 \times 10^{3}$	$3{\cdot}162 \times 10^{6}$
$3{\cdot}162 \times 10^{-4}$	$1{\cdot}000 \times 10^{-7}$	**70**	$3{\cdot}162 \times 10^{3}$	$1{\cdot}000 \times 10^{7}$
$1{\cdot}778 \times 10^{-4}$	$3{\cdot}162 \times 10^{-8}$	**75**	$5{\cdot}623 \times 10^{3}$	$3{\cdot}162 \times 10^{7}$
$1{\cdot}000 \times 10^{-4}$	$1{\cdot}000 \times 10^{-8}$	**80**	$1{\cdot}000 \times 10^{4}$	$1{\cdot}000 \times 10^{8}$
$5{\cdot}623 \times 10^{-5}$	$3{\cdot}162 \times 10^{-9}$	**85**	$1{\cdot}778 \times 10^{4}$	$3{\cdot}162 \times 10^{8}$
$3{\cdot}162 \times 10^{-5}$	$1{\cdot}000 \times 10^{-9}$	**90**	$3{\cdot}162 \times 10^{4}$	$1{\cdot}000 \times 10^{9}$
$1{\cdot}000 \times 10^{-5}$	$1{\cdot}000 \times 10^{-10}$	**100**	$1{\cdot}000 \times 10^{5}$	$1{\cdot}000 \times 10^{10}$
$3{\cdot}162 \times 10^{-6}$	$1{\cdot}000 \times 10^{-11}$	**110**	$3{\cdot}162 \times 10^{5}$	$1{\cdot}000 \times 10^{11}$
$1{\cdot}000 \times 10^{-6}$	$1{\cdot}000 \times 10^{-12}$	**120**	$1{\cdot}000 \times 10^{6}$	$1{\cdot}000 \times 10^{12}$
$3{\cdot}162 \times 10^{-7}$	$1{\cdot}000 \times 10^{-13}$	**130**	$3{\cdot}162 \times 10^{6}$	$1{\cdot}000 \times 10^{13}$
$1{\cdot}000 \times 10^{-7}$	$1{\cdot}000 \times 10^{-14}$	**140**	$1{\cdot}000 \times 10^{7}$	$1{\cdot}000 \times 10^{1}$

Section 3

Technical description

3.1 SYSTEM OPERATION

General operation of the TF 2300 Modulation Meter can be explained with reference to the block schematic, Fig. 3.1.

The r.f. input is heterodyned in the mixer with the local oscillator output, producing an intermediate frequency signal of 1.5 Mc/s. The output of the mixer is fed to an i.f. amplifier which has a linear phase/frequency response to f.m.

From the i.f. amplifier, the signal can be passed through the f.m. or the a.m. sections, depending on the positioning of the Function switch.

F.M. sections

From the i.f. amplifier, f.m. signals, as illustrated in Fig. 3.2 (a), are passed through three limiting stages to eliminate all amplitude changes and produce a rectangular waveform as shown in (b).

This limited i.f. waveform is passed to the pulse counter discriminator whose operation is illustrated in (c), (d) and (e). A pulse of fixed amplitude and width is generated every time the clipped i.f. signal passes through zero in the positive-going sense as shown in (b) and (c). At any given repetition frequency, these pulses have a constant mean amplitude, Vm, provided the pulse amplitude and width are fixed; thus when the p.r.f. varies due to f.m. of the input, the mean amplitude will also vary directly as the modulation frequency. This is illustrated in (c) and (d). In practice the limiter output is fed to a Schmitt trigger circuit, the resultant constant rise-time rectangular waveform being differentiated and used to drive a pulse generator. The pulses from here are later passed through a low-pass filter to remove all but the modulation frequency components.

The l.f. signal (e) is then amplified in the 1st l.f. amplifier, the gain of which can be standardized by the SET CAL-F.M. front panel preset in conjunction with the calibrator, and passed through the 150 kc/s low-pass filter and, if required to restrict the bandwidth to the audio range, through the 15 kc/s low-pass filter.

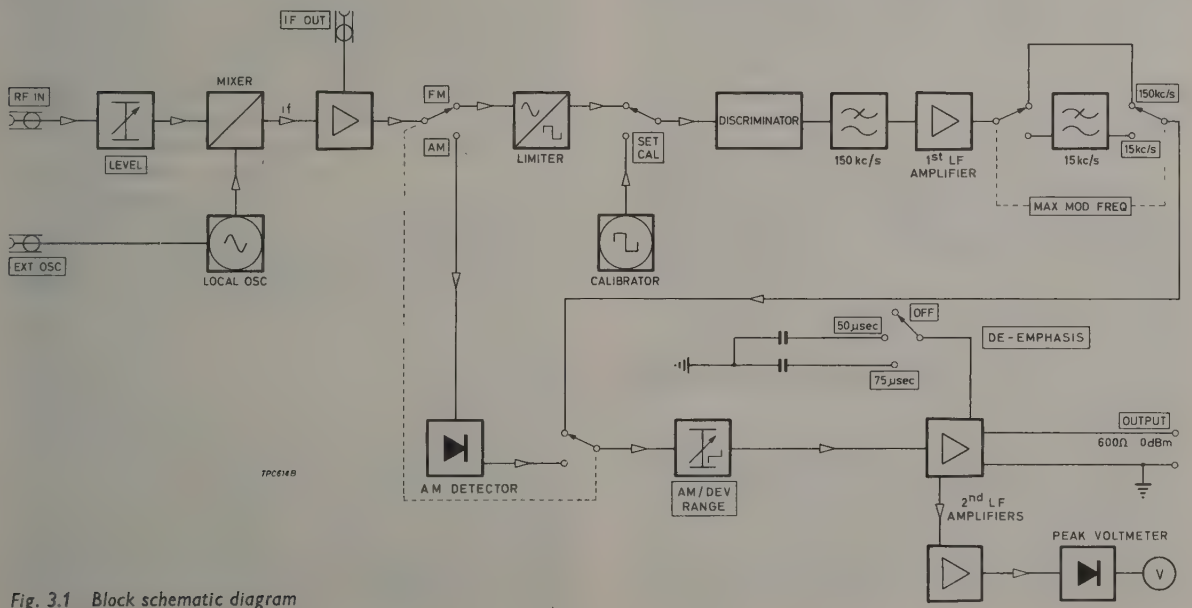


Fig. 3.1 Block schematic diagram

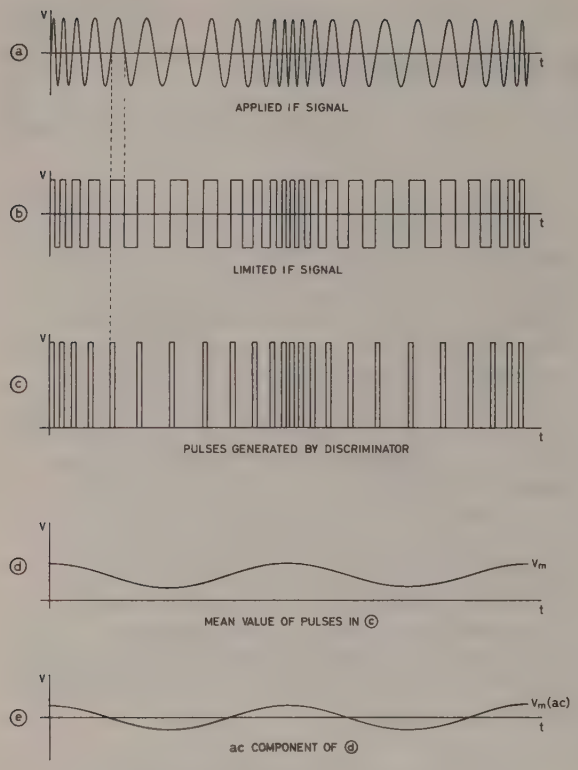


Fig. 3.2 Operation of discriminator

From this point onwards, all sections are common to f. m. and a. m. and it is necessary to return to the i. f. amplifier and consider the a. m. sections of the modulation meter before proceeding to describe these final sections.

A.M. sections

From the i. f. amplifier, a. m. signals are passed in the appropriate position of the Function switch to the a. m. detector. A diode detector is employed, preceded by an i. f. amplifier stage, the gain of which is variable by the SET A. M. control to standardize the carrier level at the detector. When the carrier level is correct, the d. c. output from the detector produces a reading at the SET line on the meter.

After detection, the signal is fed via a 50 kc/s low-pass filter to the A. M. / DEV RANGE switch and other common a. m. / f. m. sections.

The d. c. component of the detected a. m. signal is fed direct to the meter in the SET A. M. and TUNE OSCILLATOR positions of the Function switch. In the former position the d. c. component, which is independent of a. m., is used as already stated to set the carrier level at the detector, while in the latter position, when the i. f. is fixed, it is used to set the mixer input to the correct level by adjustment of the LEVEL attenuator and also for oscillator peaking adjustment during initial setting up of the modulation meter.

A.M./F.M. sections

The A. M. / DEV RANGE attenuator has a total attenuation of 40 dB switched in 10 dB steps and selects the deviation and modulation depth ranges on f. m. and a. m. On a. m., only the 10 and 20 dB steps are used.

After attenuation, the signals are passed to the 2nd l. f. amplifier which contains an output amplifier arranged to supply approximately 0 dBm into 600 Ω to the OUTPUT terminals. This stage effectively isolates the meter circuits from the OUTPUT terminals and permits de-emphasis networks to be switched in to the output terminals when required, without affecting the meter reading. The peak reading meter circuit consists of an amplifier with push-pull pair to operate the meter diode in its most linear region.

Calibration circuit

In the SET CAL position of the Function switch, the calibrator produces a standard crystal controlled deviation signal, i. e., it produces a similar signal to that from the limiter, with a peak-to-peak amplitude corresponding to a ±200 kc/s deviation. For a more complete description of the calibrator, refer to section 3. 8.

It thus provides a means of standardizing the discriminator and l. f. circuitry in order that accuracy can be maintained and also a means of checking for all possible sources of drift.

3.2 POWER UNIT *(Unit A13—Fig. 7.6)*

Occupying the rear corner behind the oscillator compartment, the power unit includes two power transistors fitted to heat sinks on the chassis, a printed board and electrolytic capacitors. One switch, SA, on the back panel selects the mains voltage range, and another, SB, the power source which may be mains or battery.

The mains input feeds straight into a low-pass filter, Z1, to prevent r.f. entering the instrument, then into a transformer T1, which has a double-wound primary. The transformer is astatically wound because of the susceptibility of the oscillator tuning system to magnetic fields. Its two primary halves are connected in series for 240 V, or parallel for 110 V, according to the position of switch SA. The secondary winding, centre tapped to earth, drives a full-wave rectifier whose output is smoothed by a conventional R-C network to give about 30 V. A series regulator follows the smoothing circuit, using the Zener diode MR5 for voltage reference. VT1, 2 and 3 form the error signal amplifier, feeding the base of the series regulator, VT6.

The output from this regulator is at -18 V, adjustable by preset potentiometer RV1, and supplies the a.m. detector, both amplifiers, the peak reading voltmeter, the discriminator and the -12 V regulator. Further isolation from mains voltage variations and hum is required for the remaining units. This is provided by the -12 V regulator, with VT4, 5 and 6 as the active elements and RV2 to set the level. The three transistors are, respectively, the error signal amplifier, emitter follower and series regulator. MR4 and VT4 are mounted in a heat sink to prevent rapid temperature differentials from affecting 12 V stability.

24 V applied to the BATTERY terminals (positive earth or isolated) passes through MR3, to avoid risk of damage through accidental polarity reversal, and goes to VT6 directly, when SB is in the BATTERY position.

3.3 MIXER *(Unit A1—Fig. 7.2)*

The mixer is a conventional, untuned, square law type, using a point contact silicon diode. It is built as a single, compact screened unit to keep spurious responses at a low level over the wide frequency range. Short signal paths are provided to allow operation near high intensity r.f. fields. The mixer is preceded by the r.f. attenuator which, at an input impedance of 50 Ω, gives 6 to 60 dB attenuation.

The r.f. input passes via the continuously variable attenuator, AT1, into the crystal mixer where mixing with the local oscillator takes place. The output from the mixer is at an i.f. of 1.5 Mc/s.

The normal input range is 4 Mc/s to 1000 Mc/s. R.F. inputs of over 1000 Mc/s can be used with reduced sensitivity. The useful upper limit is governed by oscillator stability and input sensitivity.

3.4 LOCAL OSCILLATOR *(Unit A2a—Fig. 7.2)*

VT1 produces 5.5 to 11 Mc/s directly for range 1 and 11 to 22 Mc/s second harmonic for range 2. Tuned by L1, one of the three variable inductors connected to the main tuning, the oscillator is evolved from the Colpitts circuit. There is no crystal control on this oscillator. C1 and L2 are provided to pre-adjust the range limits of frequency to the scale of the oscillator. VT1 is switched off by the switch SB on ranges 3-4 and 5-8.

Ranges 3 and 4 use VT2, tuned by L3, for the oscillator frequency, which is basically 22 to 44 Mc/s with 44 to 88 Mc/s second harmonic. The fundamental frequency is used for range 3, and the harmonic for range 4. This oscillator can be controlled by switching one of three crystals X1, 2, 3 into the circuit of VT2 in place of the de-coupling capacitor, L3 still needing to be tuned for maximum output.

Ranges 5 to 8 are derived from the 22 to 44 Mc/s oscillator, but the output goes to the multiplier board before it reaches the mixer.

Due to the permeability tuning employed in both oscillators, outputs cannot be taken from the inductor coils. The outputs are therefore taken by tapping the preset tuning capacitors, C1 and C14, by two capacitive matching systems - C4, 5 or C10, 11 in series. These outputs are fed via simple fixed attenuators to switch SB2B except in the case of VT2 on ranges 5-8, when it is fed via the doubler and harmonic generator board (A2b).

Doubler and harmonic generator *(Unit A2b—Fig. 7.2)*

VT1 is tuned, rather flatly, by L2 to act as a frequency doubler to the output from the oscillator VT2 (on A2a). L2 tunes over the range 44 to 88 Mc/s, and this signal is applied to the harmonic generator VT2 and VT3 (on A2b), whose output contains all the required harmonics of the oscillator fundamental.

3.5 I.F. AMPLIFIER *(Unit A3—Fig. 7.2)*

This board, in the narrow box alongside the local oscillator and power unit, contains three amplifying stages, each of two transistors, VT1 to VT6, the stages being coupled via band pass filters. No limiting occurs in the amplifier, and linearity, frequency response and an overall gain of 50 dB are stabilized by a negative feedback loop in each stage.

The output of the i.f. amplifier is taken in parallel paths to the Function switch, SB, and to the I. F. OUT socket via resistor R1 where it is available for viewing on an oscilloscope or for counting to check carrier frequency drift. Via SB1F, the output is routed to the a. m. detector or, in the f. m. positions, to the limiter.

3.6 LIMITER *(Unit A4—Fig. 7.3)*

Housed in the rear half of a compartment underneath the centre of the chassis, the limiter consists of three stages of emitter-coupled amplifiers, arranged so that signal amplitudes of either polarity above a certain level are limited. The emitters are connected to balancing potentiometers to equalize the excursion in each direction.

Part of the i.f. signal is tapped off from the third stage, differentiated by R15 and L1 and detected by MR1 to produce the uni-directional pulses shown in Fig. 3.3 (c). These pulses have a d.c. term, Vm, almost proportional to the repetition frequency. This d.c. is fed, in the SET FREQ positions of the Function switch, directly to the meter which is arranged to indicate SET when the i.f. is correct.

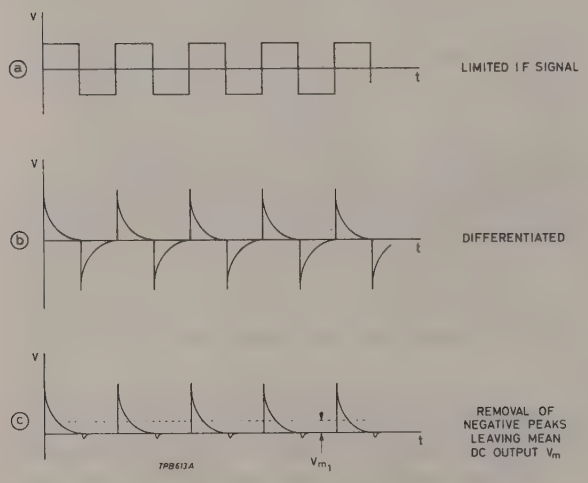


Fig. 3.3 Operation of limiter

The presence of any appreciable degree of carrier shift when modulation is applied to the input signal will be indicated by the meter in the SET FREQ positions; the amount of shift may be measured with a counter at I. F. OUT provided that the gate time is long with respect to the period of the modulating signal or is equal to an integral number of periods.

3.7 DISCRIMINATOR *(Unit A5—Fig. 7.3)*

The pulse counter type discriminator occupies the front half of the central compartment fitted beneath the chassis, the rear half of this compartment being occupied by the limiter. The overall operation of the discriminator is described in section 3.1 and illustrated in Fig. 3.2 (c), (d) and (e).

The limiter output drives a Schmitt trigger circuit, VT2 and VT3, to produce a large square wave output into C4 with constant rise and fall times. The collector voltage of VT1, the trigger amplifier, is set by RV1 so that the Schmitt circuit is on the point of regeneration.

The square wave is differentiated and passed to a pulse generator, VT4 and VT5, which produces positive-going pulses. These pulses are clipped by VT6. The emitter of this semiconductor is taken to the -12 V line, so that it clips the bases of the positive-going pulses, thus maintaining constant amplitude.

VT7 is an emitter follower which, unlike VT6, is conducting continuously and presents a constant low impedance to the 150 kc/s low-pass filter. The low frequency change in the mean value of the pulses is therefore passed and the i.f. signal rejected.

3.8 CALIBRATOR *(Unit A12—Fig. 7.5)*

VT1 and VT2 form a multivibrator running at nominally 4 kc/s. C6, R8 and VT4 differentiate and clip the negative spikes of the square wave output from VT2 and feed them from a low impedance to the binary divider, VT7 and VT8.

Since the multivibrator runs at constant frequency, the positive-going pulses from VT4 are at equal time intervals. VT7 and VT8 divide these 4 kc/s pulses by two and produce an exactly 1:1 square wave output, via C11, at about 2 kc/s. VT5 and VT6 operate as a gating circuit controlled at 2 kc/s by this square wave.

VT3 is an oscillator accurately controlled at 400 kc/s by crystal XL1 and gated on and off by VT5 and VT6 at 2 kc/s. Thus the 400 kc/s signal is divided into bursts of r.f. at a repetition frequency of 2 kc/s and a 1:1 on/off ratio - see Fig. 3.4. This is fed to the discriminator via C9 when SB is at position SET CAL and corresponds, for calibration purposes, to a crystal controlled deviation of ±200 kc/s. It should be noted that, in this mode, the 15 kc/s filter is always in circuit to reduce the tendency of high frequency ringing due to the use of a square wave for calibration.

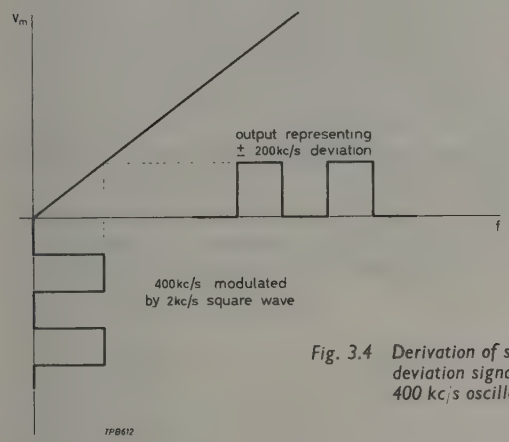


Fig. 3.4 Derivation of standard deviation signal from 400 kc/s oscillator

The calibrator operates only in the SET CAL position of the Function switch, being switched off at all other positions. A single attenuator pad contains the variable resistor A0RV3 (see Fig. 7.1). This resistor (fitted immediately behind the calibrator board on the chassis) can be used to standardize the meter deflection to the SET mark.

3.9 LOW-PASS FILTERS *(Units A6 and A8—Fig. 7.3)*

150 kc/s and 15 kc/s

These filters are 3 dB down at 250 and 28 kc/s respectively to obtain optimum flatness over their pass bands of 150 and 15 kc/s. In order to accommodate the channel separation requirements of stereo broadcast monitoring, the filter designs are such as to provide a linear phase frequency response.

The 150 kc/s filter integrates the pulses from the discriminator and rejects the i.f. signal; thus it produces a low frequency voltage which is proportional in amplitude to the deviation. The 15 kc/s filter restricts the bandwidth to the audio range and can be switched in as required by switch SE.

3.10 1st L.F. AMPLIFIER *(Unit A7—Fig. 7.3)*

Demodulated signals from the discriminator reach the first l.f. amplifier via the 150 kc/s low-pass filter. The amplifier has three stages, VT1 and 2 being a current feedback pair, while VT3 and 4 are arranged as a special low impedance configuration known as an 'emitter-squared follower'.

The amplifier has been designed to give high stability, large bandwidth and constant gain with immunity from h.t. changes, by virtue of its feedback loops. C6 carries positive feedback over part of the circuit to increase the gain, while R8 carries d.c. bias to the first stage. C5 and R7 take a.c. negative feedback to the first stage via the SET CAL F.M. control. The output stage, VT3 and VT4, gives a very low impedance, so that the output impedance of the amplifier is due almost entirely to R14. A high stability resistor is used here to give optimum matching to the following switched 15 kc/s filter. Similar considerations apply to the input stage, where the impedance is effectively lowered by parallel negative feedback, and R2 matches the preceding unit, the 150 kc/s low-pass filter.

Range attenuator *(Chassis A0—Fig. 7.1)*

Between the first and second i.f. amplifiers is the range attenuator. The 5 kc/s range is directly connected, but for each succeeding range 10 dB attenuation is switched into circuit. Two meter scales are used alternately to give 10 dB steps.

3.11 2nd L.F. AMPLIFIER *(Unit A9—Fig. 7.4)*

The action of VT1 and VT2 in this unit is similar to that of the first two transistors in the first l.f. amplifier. RV1 in the feedback loop sets the gain. Two connections are made after

VT2; from terminal 7 to the meter circuit, and from terminal 9 to the DE-EMPHASIS switch.

The remainder of the amplifier is the same as the first l.f. amplifier and its output is taken to the front panel terminals. Headphones may be used here, or a distortion factor meter, external meter (or oscilloscope). The output impedance is 600 Ω, and is isolated by C11. Do not allow d.c. to reach the OUTPUT terminals so as to avoid damaging C11, which is a low voltage electrolytic capacitor.

3.12 PEAK READING METER *(Unit A10—Fig. 7.4)*

For reading modulation depth and deviation, and for setting the f.m. calibration, the peak reading meter circuit is brought into use.

VT1 and 2 amplify the signal from the 2nd l.f. amplifier and RV1 presets the gain of the amplifier. The remainder of the circuit consists of two similar pairs of semiconductors arranged in push-pull, driving the meter via rectifier MR1. The polarity of the meter can be switched to read trough or peak on a.m., the positive or negative deviation on f.m. Feedback is used to stabilize the gain, reduce distortion and broaden the frequency response. The feedback loops include R3, R12, C14, R16, R19 and R24.

The meter is also used to measure the d.c. component of the output from the a.m. detector for i.f. level monitoring and tuning, and in the position SET FREQ it reads the voltage developed by a counter circuit in the limiter thus monitoring the frequency from the mixer for fine tuning purposes.

3.13 A.M. detector *(Unit A11—Fig. 7.5)*

The diode detector, MR1, is preceded by VT2 and VT3 which are arranged in a composite emitter follower configuration to give a large peak-to-peak signal without distortion, and thus allowing MR1 to operate over the linear part of its forward characteristic. The input to the detector is via a common-emitter amplifier, VT1, stabilized by negative feedback through R6 and R7.

MR1 is fed directly from a tuned circuit, C4 and L1, which is heavily damped by the low output impedance of VT3. This gives the circuit a wide band response which is flat to within the limits of the i.f. amplifier. MR1 feeds into a resistive load and a fraction of the voltage developed across the load is tapped off and fed to a 50 kc/s low-pass filter.

When the Function switch, SB, is in positions TROUGH or PEAK, the detected signal is fed via terminal 12 to the range switch, the 2nd i.f. amplifier and the meter circuit.

When making an a.m. measurement, the internal meter is switched directly to terminal 7 by the Function switch, SB, and the d.c. component adjusted by RV1, SET AM control, until the meter reads SET.

The d.c. component is also used for setting the input level to the i.f. amplifier, thus ensuring correct loading and good signal-to-noise ratio. On position TUNE OSCILLATOR AND ADJUST LEVEL of the Function switch, a fixed resistor, R1, is switched in place of RV1 and the correct level set up by adjusting the input attenuator until the meter indicates within the black arc.

3.14 BETWEEN-UNITS CIRCUITRY *(Unit A0—Fig. 7.1)*

The A.M./DEV RANGE switch, SC, controls three π pads of 10, 20 and 20 dB respectively. These attenuating pads are inserted in 10 dB steps as follows :

	Range	Attenuation
(1)	5 kc/s f.m.	0
(2)	15 kc/s f.m., 30% a.m.	10 dB
(3)	50 kc/s f.m., 100% a.m.	20 dB
(4)	150 kc/s f.m.	10 dB + 20 dB
(5)	500 kc/s f.m.	20 dB + 20 dB

It should be noted that the meter has two deviation scales; ranges 1, 3, 5 are associated with the bottom scale and the other ranges with the upper.

Also on the chassis is a single attenuator pad for the calibrator. This is switched in place of the range attenuator on the SET CAL position of the Function switch. This pad incorporates a variable resistor, RV3, which is mounted on the top of the chassis immediately to the rear of the calibrator board. RV3 standardizes the calibration circuit to the meter.

Section 4

Maintenance

4.1 INTRODUCTION

This section is for guidance in servicing the instrument. In the case of any difficulty, please write to or phone the Marconi Instruments Service Division (see address on back cover), quoting the type and serial numbers which are on the data plate at the rear of the instrument. If the instrument is being returned for repair, please indicate clearly the nature of the fault or the work you require to be done.

This instrument uses semiconductor devices which, although having inherent long term reliability and mechanical ruggedness, are susceptible to damage by overloading, reversed polarity and excessive heat or radiation. Avoid hazards such as reversal of batteries, prolonged soldering, strong r.f. fields or other forms of radiation, use of insulation testers or accidentally applied short circuits. Even the leakage current from an unearthed soldering iron could cause trouble. Before shorting or breaking any circuit, refer to the circuit diagrams to establish the effect on bias arrangements of the transistors.

Avoid putting the instrument above or close to hot objects, e.g., transmitters etc., where the ambient temperature may affect transistor operation.

4.2 FUSE REPLACEMENT

Two fuses are fitted in 20 x 50 mm fuse holders in the rear panel.

Ref	*Rating*	*M.I. code*
FS1	100 mA for 230 V mains	23411-052
	160 mA for 110 V mains	23411-054
FS2	500 mA	23411-005

FS2 is of the quick-blow type. This fuse must never be changed without first switching off the mains supply, or circuit damage may result.

4.3 ACCESS AND LAYOUT

Figs. 4.1 and 4.2 show the general locations of all sub-units on the chassis. When required, units may be removed or access obtained for adjustment as follows. (Refitting is generally the reverse procedure unless otherwise stated.)

Removal of case or dust covers

The case is slid off backwards after removing four screws in the rear panel. Rack-mounted versions have dust covers top and bottom, each attached by four screws.

R.F. attenuator and mixer

It is not advisable to attempt any removal of these items, the former being a sealed unit. In all cases, contact Marconi Instruments if rectification appears necessary in this unit.

Oscillator

Disconnect the plug to the mixer unit, withdraw the lead and plug the oscillator compartment. Remove the chrome screws at each corner of the oscillator front panel and slide the unit out. To remove completely, disconnect the supply plug from within the oscillator compartment. When withdrawn, all presets, coils etc., are accessible without further dismantling.

Printed circuit boards

The following are plug-in circuit boards which, apart from A13, can be removed by pressing a spring clip and pulling the board upwards :

(1) 1st l.f. amplifier (A7)

(2) 2nd l.f. amplifier (A9)

(3) A. M. detector (A11)

(4) Peak reading voltmeter (A10)

(5) Calibrator (A12)

(6) Power supply unit (A13)

(This board is secured by a bracket which has to be removed before the board can be unplugged.)

An extension board, type TM 7926, is stowed on the inside of the rear panel. By means of this board, any of the above can be raised above its compartment for examination while in the operative condition.

I.F. amplifier

The complete amplifier can be removed when necessary by unsoldering one lead and disconnecting two miniature plugs. Three screws retain the unit to the chassis. Access for adjustment is provided by a removable cover, thus enabling the inductors of the tuned circuits to be adjusted through holes in the back of the circuit boards with a right angled trimmer tool.

Limiter and discriminator

All thirteen connections are soldered; the complete unit can be removed after these have been unsoldered by releasing four nuts from the top side of the chassis. The cover is removable to give access to all the circuits and presets.

150 kc/s and 15 kc/s low-pass filters

Either of these units can be removed by unsoldering two connections and releasing two screws. The 150 kc/s filter is adjustable through holes in the top chassis, (immediately below extension board stowage). The cover of the 15 kc/s filter must be removed if adjustment is required.

4.4 CLEANING AND LUBRICATION

(1) Rotary switch contacts. Cleaning or lubrication is necessary once or twice yearly depending on usage; use a mixture of benzine and light oil or one of the proprietary preparations such as Electrolube No. 1, manufactured by Electrolube Ltd., Slough, Bucks., England.

(2) Oscillator. Excessive lubrication must be avoided but the moving parts should be lubricated and cleaned at least twice a year :

(a) Worm shaft : Use Rocol anti-scuffing paste.

(b) Carriage slide : Use Aeroshell 4 oil.

(c) Scale drum mounting : Apply a few drops of Aeroshell 4 oil to the felt washer on the spigot mounting.

(d) Range switch bevel gears : Use Rocol anti-scuffing paste.

Note : Sealed bearings at end of worm shaft will not require lubrication within the life of the instrument.

4.5 REPLACING DRIVE CORD

The oscillator drive cord consists of a $3\frac{1}{2}$ ft length of nylon cord arranged as shown in Fig. 4.3. To fit a new cord :-

(1) Turn the OSCILLATOR TUNE control fully clockwise.

(2) Attach one end of the cord (X) to the correct spring on pulley C.

(3) Pass the cord through the slot and once clockwise round pulley C, then over pulley A and counter-clockwise for three-quarters of a turn round the bush to point X'.

(4) Ease the pin in the bush clear of the hole by means of a small screwdriver. Loop the cord into the hole and push back the pin to anchor it.

(5) Starting from point Y, wind the cord $7\frac{1}{2}$ times counter-clockwise round the bush,

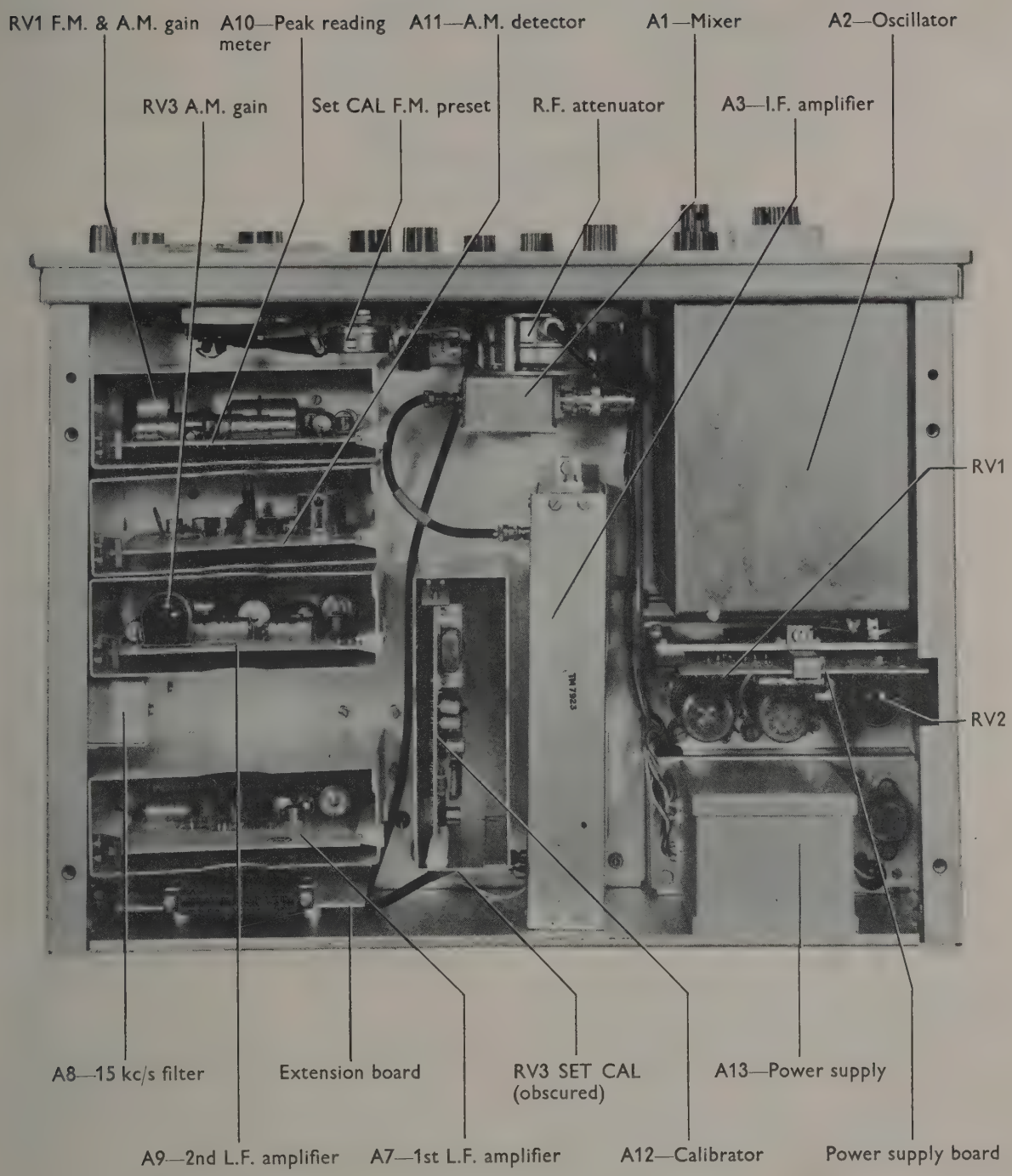


Fig. 4.1 Locations of sub-units (view underneath chassis)

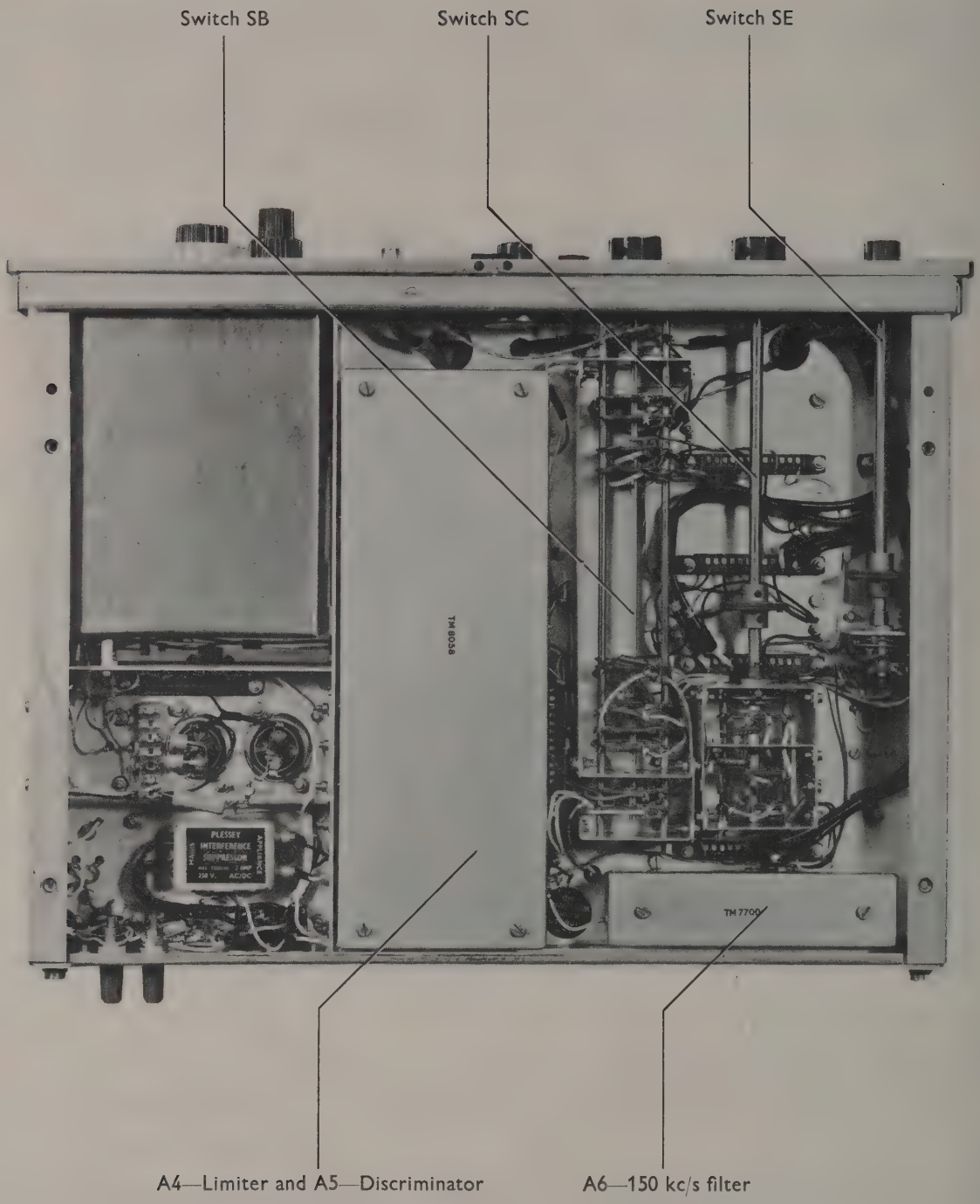


Fig. 4.2 Locations of sub-units (top view on chassis)

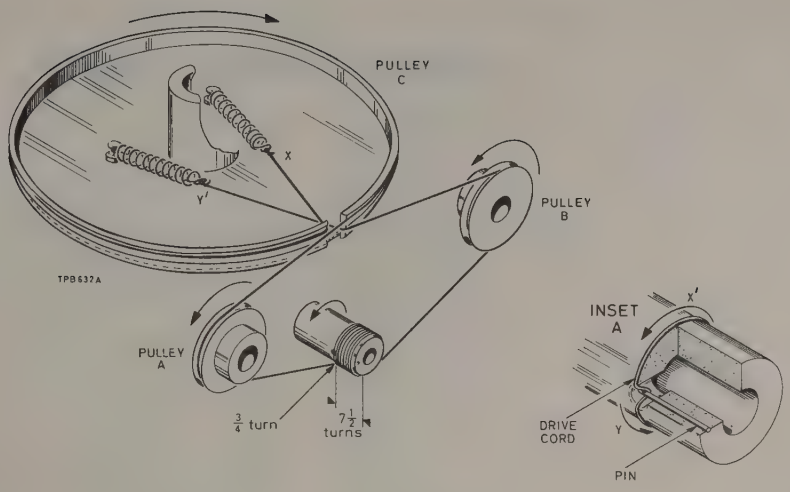


Fig. 4.3 Fitting oscillator drive cord

take it over pulley B, through the slot in pulley C and attach it to the other spring at point Y'.

4.6 CHECKING METER CALIBRATION

F.M. deviation scale

The internal f. m. calibrator provides a standard of good long-term reliability for checking and resetting the deviation reading accuracy. The calibrator accuracy is largely dependent on the frequency of its 200 kc/s oscillator crystal which is unlikely to change significantly but, if its accuracy is suspected, the deviation reading should be checked by making a normal measurement on a symmetrically modulated signal of accurately known deviation.

If the reading differs from the correct value when switched to DEV + or DEV -, adjust the SET CAL FM preset to produce the correct reading.

Then turn the Function switch to SET CAL and, without disturbing the SET CAL FM preset, bring the meter reading to the SET line by adjusting the internal f. m. preset, (A0)RV3.

A.M. depth scale

A. M. depth accuracy, unlike f. m. deviation, cannot be checked against an internal calibrating circuit. Therefore, the a. m. measurement accuracy can only be checked by making a normal measurement on a signal of accurately known modulation depth. A depth of about 50% is recommended. It is important that this signal should have a very low envelope distortion as this can cause an apparent error in depth indication.

If the reading differs from the correct value when switched to PEAK or TROUGH, correct the reading by adjusting the internal preset control (A9)RV1.

Note : Adjustment of RV1 may affect the accuracy of the f. m. deviation reading; therefore, after adjusting RV1, reset the f. m. calibration as described in section 2. 5 (not by adjusting the internal preset).

Section 6

Replaceable parts

Introduction

The modulation meter consists of a number of sub-assemblies, each of which is allocated an identity number in the sequence A0 to A13 inclusive, as follows :-

A0 - Main chassis
A1 - Mixer
A2 - Oscillator assembly
  A2a Oscillator
  A2b Doubler and harmonic generator
A3 - I.F. amplifier
A4 - Limiter
A5 - Discriminator
A6 - 150 kc/s l.p. filter
A7 - 1 st l.f. amplifier
A8 - 15 kc/s filter
A9 - 2 nd l.f. amplifier
A10 - Peak reading voltmeter
A11 - A.M. detector
A12 - Calibrator
A13 - Power unit assembly

The complete reference of a part consists of the sub-assembly number followed by its circuit reference, e.g., A4C5, A7R4, etc., and this should always be stated on any order, letter, etc. For convenience in the text and on circuit diagrams, the circuit reference is abbreviated by dropping the sub-assembly prefix except where there is risk of ambiguity.

Replaceable parts are grouped in order of sub-assembly designation and itemized in alpha-numerical order of circuit references, with miscellaneous parts at the end of each list. The following abbreviations are used :-

C : capacitor
Carb : carbon
Cer : ceramic
CRT : cathode ray tube
Elec : electrolytic
FS : fuse
ILP : illumination lamp
JK : jack
L : inductor
Log : logarithmic law
M : meter
Met : metal
Min : minimum value
MR : semiconductor diode
Ox : oxide
PL : plug
Plas : plastic
PLP : pilot lamp
R : resistor
RV : variable resistor
S : switch
SKT : socket
T : transformer
TE : total excursion
TH : thermistor
TP : terminal
V : valve
Var : variable or preset
VT : transistor
WW : wirewound
X : ferrite bead
† : value selected during test; nominal value listed
ø : feed-through component
W* : watts at 70°C
W** : watts at 55°C
W*** : watts at 40°C
W° : watts at unspecified temperature

Ordering

When ordering replacement or spare parts, address the order to our Service Division (for address see rear cover) or nearest representative. Please specify the following information for each part required.

(1) Type and serial number of instrument (see rear data plate).
(2) Complete circuit reference (see Introductio
(3) Description.
(4) M.I. code number.

If a part is not listed, state its function, location and description when ordering.

Additional component description

Detailed information on capacitors, additional to that given in the Description column of the parts lists, is included in Table 6.1.

TABLE 6.1
Capacitor details

M.I. code	Component type
26174-101/199	Metallized, tubular, plastic case.
26258-101/399	Silvered, rectangular moulded case.
26324-001/999	Low K, tubular case, epitex insulated.
26372-601/619	High K, lead-through.
26383-001/699	Disk type, epitex insulated.
26412-101/449	Etched foil, tubular metal case.
26414-101/359	Etched foil, tubular metal case.
26417-001/729	Etched foil, tubular metal case.
26427-301/399	Fabricated plate, cylindrical metal case, plug-in mounting, printed circuit.
26485-251/299	Tantalum pellet, plastic case, slide wire terminations.
26516-001/899	Tubular case.
26518-001/819	Unprotected.
26582-201/249	Metal rectangular, moulded case
26817-201/239	Var., trimmer type, solid electrode, printed circuit.

A0—MAIN ASSEMBLY, TF 2300

When ordering, prefix circuit reference with A0.

Circuit reference	Description	M.I. code
C1	Cer 0.1 μF +50-25% 25 V	26383-031
C2	Cer 0.1 μF +50-25% 25 V	26383-031
C3	Cer 0.1 μF +50-25% 25 V	26383-031
C4	Cer 0.1 μF +50-25% 25 V	26383-031
C5	Plas 0.201 μF +50-25% 25 V	26518-330
C6	Plas 0.302 μF ±1% 50 V	26518-341
C7	Plas 0.047 μF ±10% 250 V	26582-206
C8	Cer 0.1 μF +50-25% 25 V	26383-031
M1	Meter 100 μA f.s.d. 1025 Ω	44572-209
PLA	R F miniature	23441-014
PLB	R F miniature	23441-014
PLC	R F miniature	23441-014

Circuit reference	Description	M.I. code
R1	Carb 10 kΩ ±10% 0.1 W*	24341-310
R2	Met film 1.222 kΩ ±0.5% $\frac{1}{4}$W*	24635-006
R3	Met film 4.95 kΩ ±0.5% $\frac{1}{4}$W*	24635-104
R4	Met film 1.222 kΩ ±0.5% $\frac{1}{4}$W*	24635-006
R6	Met ox 22 Ω ±7% TE 3/8W**	24552-028
R8	Met film 1.925 kΩ ±0.5% $\frac{1}{4}$W*	24635-008
R9	Met film 1.423 kΩ ±0.5% $\frac{1}{4}$W*	24635-009
R10	Met film 1.925 kΩ ±0.5% $\frac{1}{4}$W*	24635-008
R11	Met film 1.222 kΩ ±0.5% $\frac{1}{4}$W*	24635-006
R12	Met film 4.95 kΩ ± 0.5% $\frac{1}{4}$W*	24635-104
R13	Met film 1.222 kΩ ±0.5% $\frac{1}{4}$W*	24635-006
R14	Met film 1.027 kΩ ±0.5% $\frac{1}{4}$W*	24635-005
R15	Met film 27.5 kΩ ±0.5% $\frac{1}{4}$W*	24635-565
R16	Met film 1.027 kΩ ±0.5% $\frac{1}{4}$W*	24635-005
R17	Met ox 100 kΩ ±7% TE 3/8 W**	24552-135

For symbols and abbreviations see introduction to this section

When ordering, prefix circuit reference with A0.

Circuit reference	Description	M.I. code
RV1	Carb 5 kΩ ±20% 2 W*	25645-374
RV2	Carb 50 kΩ ±20% 2 W*	25645-421
RV3	WW 5 kΩ ±10% 1 W°	25815-129
SKTA	50 Ω, BNC, I.F. OUT	23443-443
SKTB	2-pin	23421-504
SA	2 sec., 2 pos., SUPPLY	44321-406
SB	11 sec., 9 pos., Function	44325-901
SC	3 sec., 5 pos., A.M./DEV RANGE	44324-414
SD	1 sec., 3 pos., DE-EMPHASIS	44322-128
SE	1 sec., 2 pos., MAX MOD FREQ	44321-131
	Round knob, LEVEL	41141-209
	Round knob, SET A.M.	41141-202
	Skirted round knob, SUPPLY	41142-209
	Bar knob, large, A.M./DEV RANGE or Function	41145-206
	Bar knob, small, MAX MOD FREQ or DE-EMPHASIS	41145-208
	Terminal post, OUTPUT, insulated	23235-176
	Terminal post, OUTPUT, earthed	23235-177
	Case foot, (1 of 4)	37573-107

A1—MIXER, TM 7723

When ordering, prefix circuit reference with A1.

Circuit reference	Description	M.I. code
AT1	50 Ω, 65 dB	44425-505
C1	Cer 33 pF ±5% 750 V	26324-822
MR1	1N21C	28348-131
R1	Carb 2.7 kΩ ±10% 0.1 W*	24341-292
R2	Carb 470 Ω ±10% 0.1 W*	24341-269
R3	Carb 56 Ω ±10% 0.1 W*	24341-240
R4	Carb 10 kΩ ±10% 0.1 W*	24341-310
SKTA	R F miniature, o/p to i.f. amp.	23441-044
SKTB	50 Ω, BNC, i/p from oscillator	23443-413
SKTC	50 Ω, BNC, RF IN	23443-503

A2—OSCILLATOR ASSEMBLY, TM 8045

When ordering, prefix circuit reference with A2.

C1	Plas 1000 pF ±2% 350 V	26516-484
C2	Cer 4700 pF 350 V	26372-615
C3	Elec 10 pF +100-20% 6 V	26414-120
PLA	Plug, 2 pin	23421-204
	Socket, CRYSTAL (1 of 3)	28313-605
SKTA	50 Ω, BNC, EXT OSC	23443-503

For symbols and abbreviations see introduction to this section

When ordering, prefix circuit reference with A2.

Circuit reference	Description	M.I. code
	Knob, CRYSTAL	41141-215
	Knob, RANGE	41141-005
	Knob, fine OSCILLATOR TUNE (plastic)	41141-203
	Knob, coarse OSCILLATOR TUNE (aluminium)	31141-706
	Lead assembly, osc. output to mixer	43122-042
	Lead assembly, ext. osc. input	43122-025
	Oscillator drive cord, $3\frac{1}{2}$ ft long	16410-604
	Transit clip for locking carriage plate	35461-106

A2a—OSCILLATOR BOARD, TM 7705

When ordering, prefix circuit reference with A2a.

C1	Var air 2.5-1.34 pF	26817-238
C2	Plas 220 pF ±2% 125 V	26516-327
C3	Plas 0.001 μF ±2% 125 V	26516-481
C4	Plas 100 pF ±2% 500 V	26516-245
C5	Cer 47 pF ±2% 750 V	26324-833
C6	Cer 0.1 μF +50-25% 25 V	26383-031
C7	Cer 0.1 μF +50-25% 25 V	26383-031
C8	Cer 0.1 μF +50-25% 25 V	26383-031
C9	Cer 2.2 pF ±0.5 pF 750 V	26324-042
C10	Cer 68 pF ±2% 750 V	26324-868
C11	Cer 22 pF ±5% 750 V	26324-807
C12	Cer 22 pF ±5% 750 V	26324-807
C13	Plas 100 pF ±2% 500 V	26516-245
C14	Var air 2.5-1.34 pF	26817-238
C15	Cer 0.1 μF +50-25% 25 V	26383-031

When ordering, prefix circuit reference with A2a.

Circuit reference	Description	M.I. code
L2	Assembly TM 8073/3	44221-404
L4	Assembly TM 8073/4	44221-012
PLA	H. T. to oscillator	23443-305
PLB	O/P to mixer	23442-305
R1	Carb 220 Ω ±10% 0.1 W*	24341-258
R2	Carb 27 Ω ±10% 0.1 W*	24341-231
R3	Carb 27 Ω ±10% 0.5 W**	24342-031
R4	Carb 220 Ω ±10% 0.1 W*	24341-258
R5	Met ox 180 Ω ±7% TE 3/8 W**	24552-056
R6	Met ox 27 kΩ ±7% TE 3/8 W**	24552-120
R7	Met ox 15 kΩ ±7% TE 3/8 W**	24552-114
R8	Met ox 15 kΩ ±7% TE 3/8 W**	24552-114
R9	Met ox 7.5 kΩ ±7% TE 3/8 W**	24552-107
R10	Met ox 180 Ω ±7% TE 3/8 W**	24552-056
R11	Carb 68 Ω ±10% 0.1 W*	24341-243
R12	Carb 27 Ω ±10% 0.5 W**	24342-031
VT1	A1827	28451-577
VT2	A1827	28451-577

A2b—DOUBLER AND HARMONIC GENERATOR BOARD, TM 7706

When ordering, prefix circuit reference with A2b.

C1	Cer 47 pF ±2% 750 V	26324-833
C2	Cer 6.8 pF ±0.5 pF 750 V	26324-072
C3	Cer 0.1 μF +50-25% 25 V	26383-031
C4	Cer 0.1 μF +50-25% 25 V	26383-031

For symbols and abbreviations see introduction to this section

When ordering, prefix circuit reference with A2b.

Circuit reference	Description	M.I. code
C5	Cer 47 pF ±2% 750 V	26324-833
C6	Cer 22 pF ±5% 750 V	26324-807
C7	Var air 2.5-1.34 pF	26817-238
C8	Plas 150 pF ±2% 500 V	26516-291
C9	Cer 0.1 μF +50-25% 25 V	26383-031
C10	Cer 0.01 μF +80-20% 100 V	26383-055
C11	Paper 0.01 μF ±10% 200 V	26174-145
C12	Cer 0.1 μF +50-25% 25 V	26383-031
C13	Cer 0.01 μF +80-20% 100 V	26383-055
C14	Plas 100 pF ±2% 500 V	26516-245
C15	Cer 0.01 μF +80-20% 100 V	26383-055
L1	Assembly TM 7928/5	44251-207
L3	Assembly TM 7928/5	44251-207
L4	Assembly TB 22722/100	35124-411
L5	Assembly TM 7928/5	44251-207
L6	Assembly TB 22722/100	35124-411
R1	Carb 27 Ω ±10% 0.5 W**	24342-031
R2	Carb 220 Ω ±10% 0.1 W*	24341-258
R3	Carb 220 Ω ±10% 0.1 W*	24341-258
R4	Carb 100 Ω ±10% 0.1 W*	24341-250
R5	Carb 470 Ω ±10% 0.1 W*	24341-269
R6	Carb 220 Ω ±10% 0.1 W*	24341-258
R7	Carb 100 Ω ±10% 0.1 W*	24341-250
R8	Carb 560 Ω ±10% 0.1 W*	24341-272
VT1	2N914	28452-136
VT2	2N914	28452-136
VT3	2N914	28452-136
	Transistor pad	28488-113

A3—I.F. AMPLIFIER, TM 7132

When ordering, prefix circuit reference with A3.

Circuit reference	Description	M.I. code
C1	Cer 0.1 μF +50-25% 25 V	26383-031
C2	Cer 0.1 μF +50-25% 25 V	26383-031
C3	Cer 0.1 μF +50-25% 25 V	26383-031
C4	Cer 0.1 μF +50-25% 25 V	26383-031
C5	Mica 352 pF ±1% 750 V	26258-141
C6	Mica 352 pF ±1% 750 V	26258-141
C7	Cer 0.1 μF +50-25% 25 V	26383-031
C8	Cer 0.1 μF +50-25% 25 V	26383-031
C9	Cer 0.1 μF +50-25% 25 V	26383-031
C10	Cer 0.1 μF +50-25% 25 V	26383-031
C11	Mica 352 pF ±1% 750 V	26258-141
C12	Mica 352 pF ±1% 750 V	26258-141
C13	Cer 0.1 μF +50-25% 25 V	26383-031
C14	Cer 0.1 μF +50-25% 25 V	26383-031
C15	Cer 0.1 μF +50-25% 25 V	26383-031
C16	Cer 0.1 μF +50-25% 25 V	26383-031
C17	Cer 0.1 μF +50-25% 25 V	26383-031
C18	Cer 0.1 μF +50-25% 25 V	26383-031
C19	Cer 0.001 μF	26372-611
L1	55.2 μH	44257-402
L2	55.2 μH	44257-402
L3	55.2 μH	44257-402
L4	55.2 μH	44257-402
L5	55.2 μH	44257-402
L6	55.2 μH	44257-402
R1	Met ox 3.9 Ω ±7% TE 3/8W**	24552-096
R2	Met ox 18 kΩ ±7% TE 3/8 W**	24552-116
R3	Met ox 820 Ω ±7% TE 3/8 W**	24552-078

For symbols and abbreviations see introduction to this section

When ordering, prefix circuit reference with A3.

Circuit reference	Description	M.I. code
R4	Met ox 240 Ω ±7% TE 3/8 W**	24552-060
R5	Met ox 3.9 Ω ±7% TE 3/8 W**	24552-096
R6	Carb 39 Ω ±10% 0.1 W*	24341-228
R7	Carb 1.5 kΩ ±10% 0.1 W*	24341-284
R8	Met ox 560 Ω ±7% TE 3/8 W**	24552-072
R9	Met ox 470 Ω ±7% TE 3/8 W**	24552-069
R10	Met ox 2.7 kΩ ±7% TE 3/8 W**	24552-092
R11	Met ox 11 kΩ ±7% TE 3/8 W**	24552-111
R12	Met ox 1 kΩ ±7% TE 3/8 W**	24552-080
R13	Met ox 2.2 kΩ ±7% TE 3/8 W**	24552-088
R14	Carb 39 Ω ±10% 0.1 W*	24341-228
R15	Carb 1.5 kΩ ±10% 0.1 W*	24341-284
R16	Met ox 560 Ω ±7% TE 3/8 W**	24552-072
R17	Met ox 470 Ω ±7% TE 3/8 W**	24552-069
R18	Met ox 18 kΩ ±7% TE 3/8 W**	24552-116
R19	Met ox 4.7 Ω ±7% TE 3/8 W**	24552-100
R20	Met ox 3.3 kΩ ±7% TE 3/8 W**	24552-094
R21	Met ox 560 Ω ±7% TE 3/8 W**	24552-072
R22	Met ox 1.2 kΩ ±7% TE 3/8 W**	24552-082
R23	Carb 22 Ω ±10% 0.1 W*	24341-235
R24	Carb 680 Ω ±10% 0.1 W*	24341-228
SKTA	R F miniature	23441-044
SKTB	R F miniature	23441-044
VT1	2G401	28422-718
VT2	2G401	28422-718
VT3	2G401	28422-718
VT4	2G401	28422-718
VT5	2G401	28422-718
VT6	2G403	28422-728

A4—LIMITER, TM 7825

When ordering, prefix circuit reference with A4.

Circuit reference	Description	M.I. code
C1	Cer 0.1 μF +50-25% 25 V	26383-031
C2	Cer 0.1 μF +50-25% 25 V	26383-031
C3	Cer 0.1 μF +50-25% 25 V	26383-031
C4	Cer 0.1 μF +50-25% 25 V	26383-031
C5	Cer 0.1 μF +50-25% 25 V	26383-031
C6	Cer 0.1 μF +50-25% 25 V	26383-031
C7	Cer 0.1 μF +50-25% 25 V	26383-031
C8	Cer 0.1 μF +50-25% 25 V	26383-031
C9	Cer 0.1 μF +50-25% 25 V	26383-031
C10	Cer 0.1 μF +50-25% 25 V	26383-031
C11	Cer 0.1 μF +50-25% 25 V	26383-031
C12	Cer 0.1 μF +50-25% 25 V	26383-031
C13	Cer 0.0047 pF	26372-615
C14	Cer 0.0047 pF	26372-615
L1	Assembly TM 8073/2	44253-208
MR1	CG85H	28321-201
R1	Carb 1 kΩ ±10% 0.1 W*	24341-280
R2	Carb 330 Ω ±10% 0.1 W*	24341-263
R3	Carb 680 Ω ±10% 0.1 W*	24341-276
R4	Carb 470 Ω ±10% 0.1 W*	24341-269
R5	Carb 100 Ω ±10% 0.1 W*	24341-250
R6	Carb 680 Ω ±10% 0.1 W*	24341-276
R7	Carb 330 Ω ±10% 0.1 W*	24341-263
R8	Carb 1 kΩ ±10% 0.1 W*	24341-280
R9	Carb 1 kΩ ±10% 0.1 W*	24341-280
R10	Carb 330 Ω ±10% 0.1 W*	24341-263

For symbols and abbreviations see introduction to this section

When ordering, prefix circuit reference with A4.

Circuit reference	Description	M.I. code
R11	Carb 470 Ω ±10% 0.1 W*	24341-269
R12	Carb 100 Ω ±10% 0.1 W*	24341-250
R13	Carb 680 Ω ±10% 0.1 W*	24341-276
R14	Carb 330 Ω ±10% 0.1 W*	24341-263
R15	Met ox 2 kΩ ±7% TE 3/8 W**	24552-087
R16	Carb 1 kΩ ±10% 0.1 W*	24341-280
R17	Carb 330 Ω ±10% 0.1 W*	24341-263
R18	Met ox 560 Ω ±7% TE 3/8 W**	24552-072
R19	Met ox 470 Ω ±7% TE 3/8 W**	24552-069
R20	Carb 100 Ω ±10% 0.1 W*	24341-250
R21	Carb 180 Ω ±10% 0.1 W*	24341-256
R22	Carb 330 Ω ±10% 0.1 W*	24341-263
R23	Met ox 2.2 kΩ ±7% TE 3/8 W**	24552-088
R24	Met ox 3.3 kΩ ±7% TE 3/8 W**	24552-094
RV1	Carb 100 Ω ±20% ¼ W**	25611-242
RV2	Carb 100 Ω ±20% ¼ W**	25611-242
RV3	Carb 100 Ω ±20% ¼ W**	25611-242
VT1	2G401	28422-718
VT2	2G401	28422-718
VT3	2G401	28422-718
VT4	2G401	28422-718
VT5	2G401	28422-718
VT6	2G401	28422-718
X1	Ferrite bead, Mullard FX1115	23635-833
X2	Ferrite bead, Mullard FX1115	23635-833
X3	Ferrite bead, Mullard FX1115	23635-833
X4	Ferrite bead, Mullard FX1115	23635-833
X5	Ferrite bead, Mullard FX1115	23635-833

A5—DISCRIMINATOR, TM 7780

When ordering, prefix circuit reference with A5.

Circuit reference	Description	M.I. code
C1	Carb 0.1 μF +50-25% 25 V	26383-031
C2	Cer 10 pF ±0.5 pF 750 V	26324-085
C3	Cer 10 pF ±0.5 pF 750 V	26324-085
C4	Cer 22 pF ±5% 750 V	26324-807
C5	Carb 0.1 μF +50-25% 25 V	26383-031
C6	Plas 30 pF ±2 pF 500 V	26516-119
C7	Elec 100 μF +100-20% 25 V	26417-158
C8	Carb 0.1 μF +50-25% 25 V	26383-031
C9	Carb 0.1 μF +50-25% 25 V	26383-031
C10	Elec 100 μF +100-20% 25 V	26417-158
C11	Carb 0.1 μF +50-25% 25 V	26383-031
C12	Carb 0.1 μF +50-25% 25 V	26383-031
C13	Cer 0.0047 μF	26372-615
C14	Cer 0.0047 μF	26372-615
MR1	CG85H	28321-201
MR2	CG85H	28321-201
R1	Carb 1.5 kΩ ±10% ½W**	24342-084
R2	Carb 1.8 kΩ ±10% ½W**	24342-086
R3	Carb 3.3 kΩ ±10% ½W**	24342-094
R4	Carb 33 Ω ±10% ½W**	24342-033
R5	Carb 1 kΩ ±10% ½W**	24342-080
R6	Carb 100 Ω ±10% ½W**	24342-050
R7	Carb 1 kΩ ±10% ½W**	24342-080
R8	Carb 10 kΩ ±10% ½W**	24342-110
R9	Carb 22 kΩ ±10% ½W**	24342-118
R10	Carb 5.6 Ω ±10% ½W**	24342-103
R11	Carb 10 kΩ ±10% ½W**	24342-110
R12	Met ox 1 kΩ ±7% TE 3/8W**	24552-080
R13	Met ox 240 Ω ±7% TE 3/8 W**	24552-060

For symbols and abbreviations see introduction to this section

When ordering, prefix circuit reference with A5.

Circuit reference	Description	M.I. code
R14	Met ox 1 kΩ ±7% TE 3/8W**	24552-080
R15	Met ox 6.8 Ω ±7% TE 3/8W**	24552-106
R16	Carb 10 Ω ±10% ½W**	24342-020
R17	Carb 470 Ω ±10% ½W**	24342-069
R18	Carb 470 Ω ±10% ½W**	24342-069
RV1	Carb 4.7 kΩ ±20% ¼W**	25611-209
VT1	2N706	28433-356
VT2	2N706	28433-356
VT3	2N706	28433-356
VT4	2N706	28433-356
VT5	BSY27	28452-843
VT6	2N706	28433-356
VT7	2N706	28433-356
X1	Ferrite bead, Mullard FX1115	23635-833
X2	Ferrite bead, Mullard FX1115	23635-833
X3	Ferrite bead, Mullard FX1115	23635-833

A6—150 kc/s FILTER, TM 7700

When ordering, prefix circuit reference with A6.

Circuit reference	Description	M.I. code
C1	Mica 720 pF ±1% 350 V	26258-150
C2	Mica 0.00144 µF ±1% 350 V	26258-185
C3	Mica 720 pF ±1% 350 V	26258-150
L1	Assembly TM 7842	44266-821
L2	Assembly TM 7842	44266-821
R1	Met film 1 kΩ ±1% ¼W**	24637-101

A7—1st L.F. AMPLIFIER, TM 7223

When ordering, prefix circuit reference with A7.

Circuit reference	Description	M.I. code
C1	Elec 500 µF +50-20% 12 V	26417-172
C2	Elec 100 µF +100-20% 25 V	26417-158
C3	Cer 15 pF ±5% 750 V	26324-795
C4	Elec 100 µF +50-20% 6 V	26417-154
C5	Elec 100 µF +100-20% 25 V	26417-158
C6	Elec 100 µF +50-20% 6 V	26417-154
C7	Elec 500 µF +50-20% 12 V	26417-172
R1	Carb 100 Ω ±10% ½W**	24342-050
R2	Met film 1 kΩ ±7% TE 3/8 W**	24637-101
R3	Carb 5.6 kΩ ±10% ½W**	24342-103
R4	Carb 2.2 kΩ ±10% ½W**	24342-088
R5	Carb 4.7 kΩ ±10% ½W**	24342-100
R6	Carb 5.6 kΩ ±10% ½W**	24342-103
R7	Met ox 6.8 kΩ ±7% TE 3/8W**	24552-106
R8	Met ox 20 kΩ ±7% TE 3/8 W**	24552-117
R9	Carb 2.2 kΩ ±10% ½W**	24342-088
R10	Carb 56 Ω ±10% ½W**	24342-040
R11	Carb 82 Ω ±10% ½W**	24342-046
R12	Carb 22 kΩ ±10% ½W**	24342-118
R13	Carb 1 kΩ ±10% ½W**	24342-080
R14	Met film 1 kΩ ±1% ¼W*	24637-101
VT1	2G401	28442-718
VT2	2G401	28442-718
VT3	2N1304	28443-528
VT4	2N1304	28443-528

For symbols and abbreviations see introduction to this section

A8—15 kc/s FILTER, TM 7702

When ordering, prefix circuit reference with A8.

Circuit reference	Description	M.I. code
C1	Plas 0.0067 μF ±1% 125 V	26516-683
C2	Plas 0.0067 μF ±1% 125 V	26516-683
L1	Assembly TM 7842/1	44271-409

A9—2nd L.F. AMPLIFIER, TM 7224

When ordering, prefix circuit reference with A9.

Circuit reference	Description	M.I. code
C1	Elec 500 μF +50-20% 12 V	26417-172
C2	Cer 0.1 μF +50-20% 25 V	26383-031
C3	Cer 15 pF ±5% 750 V d.c.	26324-795
C4	Elec 100 μF +100-20% 25 V	26417-158
C5	Elec 100 μF +50-20% 6 V	26417-154
C6	Elec 100 μF +50-20% 6 V	26417-154
C7	Elec 500 μF +50-20% 12 V	26417-172
C8	Cer 15 pF ±5% 750 V d.c.	26324-795
C9	Elec 100 μF +50-20% 6 V	26417-154
C10	Elec 100 μF +50-20% 6 V	26417-154
C11	Elec 500 μF +50-20% 12 V	26417-172
C12	Elec 100 μF +100-20% 25 V	26417-158
R1	Carb 47 Ω ±10% ½W**	24342-037
R2	Met film 1 kΩ ±1% ¼W*	24637-101
R3	Carb 5.6 kΩ ±10% ½W**	24342-103

When ordering, prefix circuit reference with A9.

Circuit reference	Description	M.I. code
R4	Carb 2.2 kΩ ±10% ½W**	24342-088
R5	Carb 4.7 kΩ ±10% ½W**	24342-100
R6	Carb 5.6 kΩ ±10% ½W**	24342-103
R7	Met ox 20 kΩ ±7% TE 3/8 W*	24552-117
R8	Met ox 10 kΩ ±7% TE 3/8 W*	24552-110
R9	Carb 2.2 kΩ ±10% ½W**	24342-088
R10	Met film 500 Ω ±1% ¼W*	24636-808
R11	Met film 500 Ω ±1% ¼W*	24636-808
R12	Carb 5.6 kΩ ±10% ½W**	24342-103
R13	Carb 2.2 kΩ ±10% ½W**	24342-088
R14	Carb 4.7 kΩ ±10% ½W**	24342-100
R15	Carb 5.6 kΩ ±10% ½W**	24342-103
R16	Met ox 20 kΩ ±7% TE 3/8 W*	24552-117
R17	Carb 2.2 kΩ ±10% ½W**	24342-088
R18	Carb 56 Ω ±10% ½W**	24342-040
R19	Carb 82 Ω ±10% ½W**	24342-046
R20	Carb 22 kΩ ±10% ½W**	24342-118
R21	Carb 1 kΩ ±10% ½W**	24342-080
R22	Met film 600 Ω ±1% ¼W*	24636-802
RV1	WW 22 kΩ ±10% 1W*	25811-554
VT1	2G401	28422-718
VT2	2G401	28422-718
VT3	2G401	28422-718
VT4	2G401	28422-718
VT5	2N1304	28443-528
VT6	2N1304	28443-528

For symbols and abbreviations see introduction to this section

A10—PEAK READING VOLTMETER, TM 7218

When ordering, prefix circuit reference with A10.

Circuit reference	Description	M.I. code
C1	Elec 100 μF +100-20% 6 V	26417-154
C2	Cer 0.1 μF +50-25% 25 V	26383-031
C3	Elec 500 μF +50-20% 12 V	26417-172
C4	Elec 100 μF +100-20% 25 V	26417-158
C5	Elec 100 μF +100-20% 6 V	26417-154
C6	Elec 100 μF +100-20% 6 V	26417-154
C7	Elec 100 μF +100-20% 25 V	26417-158
C8	Elec 15 μF ±20% 15 V	26485-278
C9	Elec 100 μF +100-20% 25 V	26417-158
C10	Elec 100 μF +100-20% 6 V	26417-154
C11	Elec 100 μF +100-20% 6 V	26417-154
C12	Elec 100 μF +100-20% 25 V	26417-158
C13	Cer 6.8 pF ±0.5 pF 750 V	26324-072
C14	Paper 0.01 μF ±10% 200 V	26174-145
MR1	1N914	28336-676
R1	Met ox 3.9 kΩ ±7% TE 3/8 W**	24552-096
R2	Carb 5.6 kΩ ±10% ½W**	24342-103
R3	Met ox 15 kΩ ±7% TE 3/8 W**	24552-114
R4	Carb 2.7 kΩ ±10% ½W**	24342-092
R5	Carb 4.7 kΩ ±10% ½W**	24342-100
R6	Carb 2.2 kΩ ±10% ½W**	24342-088
R7	Met ox 1.2 kΩ ±7% TE 3/8 W**	24552-082
R8	Carb 10 kΩ ±10% ½W*	24342-110
R9	Carb 2.2 kΩ ±10% ½W**	24342-088
R10	Carb 5.6 kΩ ±10% ½W**	24342-103
R11	Carb 4.7 kΩ ±10% ½W**	24342-100
R12	Met ox 20 kΩ ±7% TE 3/8 W**	24552-117
R13	Met ox 470 Ω ±7% TE 3/8 W**	24552-069
R14	Met ox 1 kΩ ±7% TE 3/8 W**	24552-080
R15	Met ox 68 kΩ ±7% TE 3/8 W**	24552-131
R16	Met ox 20 kΩ ±7% TE 3/8 W**	24552-117
R17	Met ox 1 kΩ ±7% TE 3/8 W**	24552-080
R18	Met ox 470 Ω ±7% TE 3/8 W**	24552-069
R19	Met ox 20 kΩ ±7% TE 3/8 W**	24552-117
R20	Carb 2.2 kΩ ±10% ½W**	24342-088
R21	Carb 5.6 kΩ ±10% ½W**	24342-103
R22	Carb 4.7 kΩ ±10% ½W**	24342-100
R23	Carb 10 kΩ ±10% ½W**	24342-110
R24	Carb 470 Ω ±10% ½W**	24342-152
RV1	WW 1.5 kΩ ±10% 1 W	25811-542
VT1	2G401	28422-718
VT2	2G401	28422-718
VT3	2G401	28422-718
VT4	2N706	28433-356
VT5	2N706	28433-356
VT6	2G401	28422-718

A11—A.M. DETECTOR, TM 7276

When ordering, prefix circuit reference with A11.

Circuit reference	Description	M.I. code
C1	Cer 0.1 μF +50-25% 25 V	26383-031
C2	Cer 0.1 μF +50-25% 25 V	26383-031
C3	Cer 0.1 μF +50-25% 25 V	26383-031
C4	Plas 680 pF ±2% 125 V	26516-444
C5	Cer 0.1 μF +50-25% 25 V	26383-031

For symbols and abbreviations see introduction to this section

When ordering, prefix circuit reference with A11.

Circuit reference	Description	M.I. code
C6	Plas 0.0028 μF ±2% 125 V	26516-591
C7	Plas 0.0028 μF ±2% 125 V	26516-591
C8	Plas 0.0028 μF ±2% 125 V	26516-591
C9	Plas 0.0028 μF ±2% 125 V	26516-591
L1	Assembly TM 8073/1	44251-004
L2	Assembly TM 7842/1	44267-403
L3	Assembly TM 7842/1	44267-403
MR1	CG85H	28321-201
R1	Met ox 2 kΩ ±7% TE 3/8 W**	24552-087
R2	Met ox 620 Ω ±7% TE 3/8 W**	24552-074
R3	Met ox 1 kΩ ±7% TE 3/8 W**	24552-080
R4	Met ox 680 Ω ±7% TE 3/8 W**	24552-076
R5	Met ox 8.2 kΩ ±7% TE 3/8 W**	24552-108
R6	Met ox 20 kΩ ±7% TE 3/8 W**	24552-117
R7	Met ox 20 kΩ ±7% TE 3/8 W**	24552-117
R8	Met ox 1.5 kΩ ±7% TE 3/8 W**	24552-084
R9	Met ox 470 Ω ±7% TE 3/8 W**	24552-069
R10	Met ox 2.7 kΩ ±7% TE 3/8 W**	24552-092
R11	Met film 12 kΩ ±1% ¼W*	24637-507
R12	Met ox 820 Ω ±7% TE 3/8 W**	24552-078
R13	Met ox 330 Ω ±7% TE 3/8 W**	24552-063
VT1	2G401	28422-718
VT2	2G403	28424-728
VT3	2N706	28433-356

A12—CALIBRATOR, TM 7620

When ordering, prefix circuit reference with A12.

Circuit reference	Description	M.I. code
C1	33 pF ±5% 750 V	26324-822
C2	Plas 0.047 μF ±10% 250 V	26582-206
C3	Cer 300 pF ±20% 600 V	26324-119
C4	Cer 300 pF ±10% 600 V	26324-119
C5	Plas 0.047 μF ±10% 250 V	26582-206
C6	Paper 0.001 μF ±10% 500 V	26174-125
C7	Plas 0.1 μF ±10% 250 V	26582-208
C8	Paper 0.001 μF ±10% 500 V	26174-125
C9	Elec 100 μF +100-20% 25 V	26417-158
C10	Paper 0.004 μF ±10% 250 V	26174-137
C11	Elec 100 μF +100-20% 25 V	26417-158
C12	Paper 0.004 μF ±10% 250 V	26174-137
C13	Paper 0.001 μF ±10% 500 V	26174-125
C14	Elec 100 μF +100-20% 25 V	26417-158
C15	Cer 0.1 μF +50-25% 25 V	26383-031
MR1	CG85H	28321-201
MR2	CG85H	28321-201
R1	Carb 4.7 kΩ ±10% ½W**	24342-100
R2	Carb 560 Ω ±10% ½W**	24342-072
R3	Carb 120 kΩ ±10% ½W**	24342-137
R4	†Carb 180 kΩ ±10% ½W**	24342-141
R5	Carb 560 Ω ±10% ½W**	24342-072
R6	Carb 4.7 kΩ ±10% ½W**	24342-100
R7	Carb 2.7 kΩ ±10% ½W**	24342-092
R8	Carb 5.6 kΩ ±10% ½W**	24342-103
R9	Carb 1 kΩ ±10% ½W**	24342-080
R10	Carb 22 kΩ ±10% ½W**	24342-118

For symbols and abbreviations see introduction to this section

When ordering, prefix circuit reference with A12.

Circuit reference	Description	M.I. code
R11	†Carb 18 kΩ ±10% ½W**	24342-116
R12	Carb 1 kΩ ±10% ½W**	24342-080
R13	Carb 15 kΩ ±10% ½W**	24342-114
R14	Carb 10 kΩ ±10% ½W**	24342-110
R15	Carb 22 kΩ ±10% ½W**	24342-118
R16	Carb 10 kΩ ±10% ½W**	24342-110
R17	Carb 10 kΩ ±10% ½W**	24342-110
R18	Carb 560 Ω ±10% ½W**	24342-072
R19	Carb 470 Ω ±10% ½W**	24342-069
R20	Carb 560 Ω ±10% ½W**	24342-072
R21	Carb 10 kΩ ±10% ½W**	24342-110
R22	Carb 15 kΩ ±10% ½W**	24342-114
R23	Carb 10 kΩ ±10% ½W**	24342-110
R24	Carb 10 Ω ±10% ½W**	24342-020
VT1	2S323	28433-318
VT2	2S323	28433-318
VT3	2N404	28423-508
VT4	2N404	28423-508
VT5	2N404	28423-508
VT6	2N404	28423-508
VT7	2N404	28423-508
VT8	2N404	28423-508
XL1	Crystal QM27U	28311-690
	Socket, McMurdo X2/UB	28313-603

A13—POWER UNIT ASSEMBLY, TM 7454

When ordering, prefix circuit reference with A13.

Circuit reference	Description	M.I. code
C1	Elec 500 μF +50-20% 70 V	26427-323
C2	Elec 500 μF +50-20% 70 V	26427-323
C3	Elec 0.2 μF +100-20% 70 V	26412-212
C4	Elec 100 μF +100-20% 25 V	26417-158
C5	Elec 250 μF +100-20% 25 V	26417-167
C6	Paper 0.1 μF ±10% 350 V	26174-173
C7	Elec 100 μF +100-20% 25 V	26417-158
C8	Elec 250 μF +100-20% 25 V	26417-167
C9	Cer 4700 pF 350 V	26372-615
C10	Cer 4700 pF 350 V	26372-615
C11	Cer 4700 pF 350 V	26372-615
FS1	Alternative 100 mA (for 230 V)	23411-052
	160 mA (for 110 V)	23411-054
FS2	500 mA	23411-005
MR1	1N540	28357-044
MR2	1N540	28357-044
MR3	1N540	28357-044
MR4	1S44	28357-548
MR5	ZB6.8	28371-546
MR6	ZB6.8	28371-546
PLA	3-pin, mains plug	23423-151
R1	Met ox 24 Ω ±5% 6 W ***	24588-402
R2	Met ox 10 kΩ ±7% 3/8 W **	24552-110
R3	Met ox 390 Ω %7% 3/8 W **	24552-065
R4	Carb 18 kΩ ±10% ¼ W^{O}	24342-116
R5	Carb 3.3 kΩ ±10% ¼ W^{O}	24342-094

For symbols and abbreviations see introduction to this section

When ordering, prefix circuit reference with A13.

Circuit reference	Description	M.I. code
R6	Met ox 1.2 kΩ ±7% 3/8 W **	24552-082
R7	Met ox 1.6 kΩ ±7% 3/8 W **	24552-085
R8	Carb 5.6 kΩ ±10% ¼ W^{O}	24342-103
R9	Carb 10 kΩ ±10% ¼ W^{O}	24342-110
R10	Carb 1.2 kΩ ±10% ¼ W^{O}	24342-082
R11	Met ox 1.2 kΩ ±7% 3/8 W **	24552-082
R12	Met ox 750 Ω ±7% 3/8 W **	24552-077
R13	Carb 10 kΩ ±10% ¼ W^{O}	24342-110
R14	Met ox 2 kΩ ±7% 3/8 W **	24552-087
RV1	Carb 1 kΩ	25611-014
RV2	WW 1 kΩ	25811-538
SA	Switch d.p.d.t. 110-230 V	23467-119
SB	Switch s.p.d.t. MAINS/ BATT	23467-116

When ordering, prefix circuit reference with A13.

Circuit reference	Description	M.I. code
T1	Assembly TM 7477/1	43456-004
Z1	Filter, mains	23642-901
VT1	2N1304	28443-528
VT2	2N404	28423-508
VT3	ACY17	28426-497
VT4	2N404	28423-508
VT5	ACY17	28426-497
	Fuse holder	23416-191
	Locking plate for switch SA	34444-116
	Battery terminal (+)	23235-177
	Battery terminal (-)	23235-176

For symbols and abbreviations see introduction to this section

Section 7

Circuit diagrams

CIRCUIT NOTES

1. ARRANGEMENT

The inter-unit wiring diagram, Fig. 7.1, shows all sub-assembly units in the equipment together with their reference designators (A1, A2 etc.) and type numbers (prefixed TM). Components that are not on a sub-assembly are part of the main chassis assembly (designated A0).

Circuit diagrams are arranged in order of the sub-assembly designations.

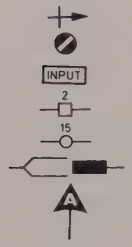
2. COMPONENT VALUES

Resistors : No suffix = ohms, k = kilohms, M = megohms.
Capacitors : No suffix = microfarads, p = picofarads.
Inductors : No suffix = henries, m = millihenries, μ = microhenries.
† : value selected during test, nominal value shown.

3. VOLTAGES

Printed in italics. Voltages are d.c. and relative to chassis unless otherwise indicated. Measured with a 20 kΩ/V meter.

4. SYMBOLS

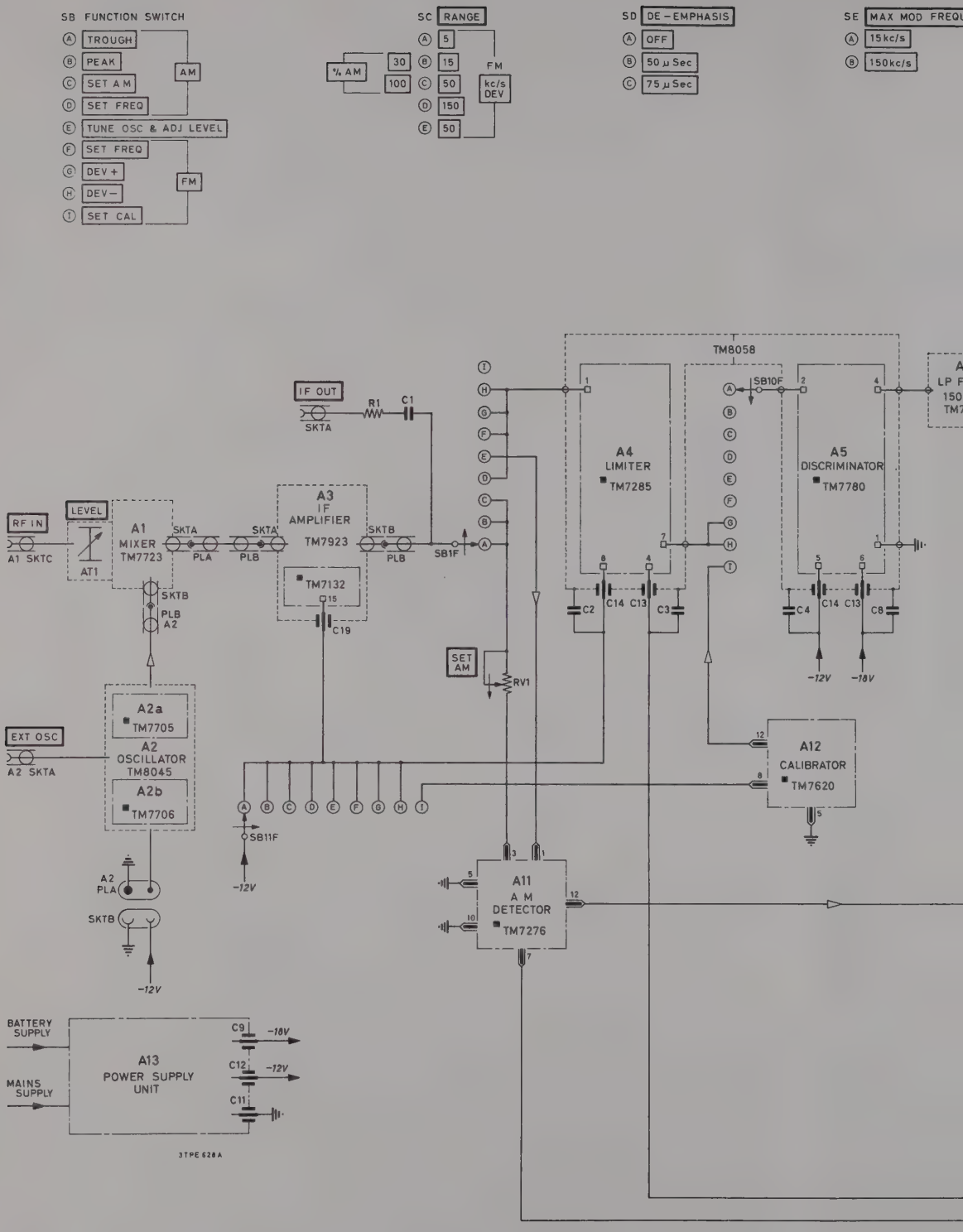


arrow indicates clockwise rotation of knob.
preset component.
panel marking.
printed board tag number.
other tag.
printed board edge connector.
indicates points at same supply potential.

5. SWITCHES

Rotary switches are drawn schematically. Numbers or letters indicate control knob setting as shown in the key diagrams. Sequence of sections reading from control knob end is as follows :-

1F = 1st section, front
1B = 1st section, back
2F = 2nd section, front
etc.


SB FUNCTION SWITCH
A TROUGH
B PEAK
C SET A M
D SET FREQ
E TUNE OSC & ADJ LEVEL
F SET FREQ
G DEV +
H DEV −
I SET CAL
AM
FM
SC RANGE
% AM
30
100
A 5
B 15
C 50
D 150
E 50
FM
kc/s DEV
SD DE−EMPHASIS
A OFF
B 50 µ Sec
C 75 µ Sec
SE MAX MOD FREQ
A 15kc/s
B 150kc/s
TM8058
IF OUT
SKTA
R1
C1
A4
LIMITER
TM7285
A5
DISCRIMINATOR
TM7780
SB10F
RF IN
A1 SKTC
LEVEL
AT1
A1
MIXER
TM7723
SKTA
PLA
PLB
A3
IF
AMPLIFIER
TM7923
SKTB
SB1F
TM7132
C19
C2
C14
C13
C3
C4
C8
−12V
−18V
SET AM
RV1
EXT OSC
A2 SKTA
A2a
TM7705
A2
OSCILLATOR
TM8045
A2b
TM7706
A2 PLA
SKTB
SB11F
A12
CALIBRATOR
TM7620
A11
A M
DETECTOR
TM7276
BATTERY SUPPLY
MAINS SUPPLY
A13
POWER SUPPLY
UNIT
C9
C12
C11
3TPE 628A

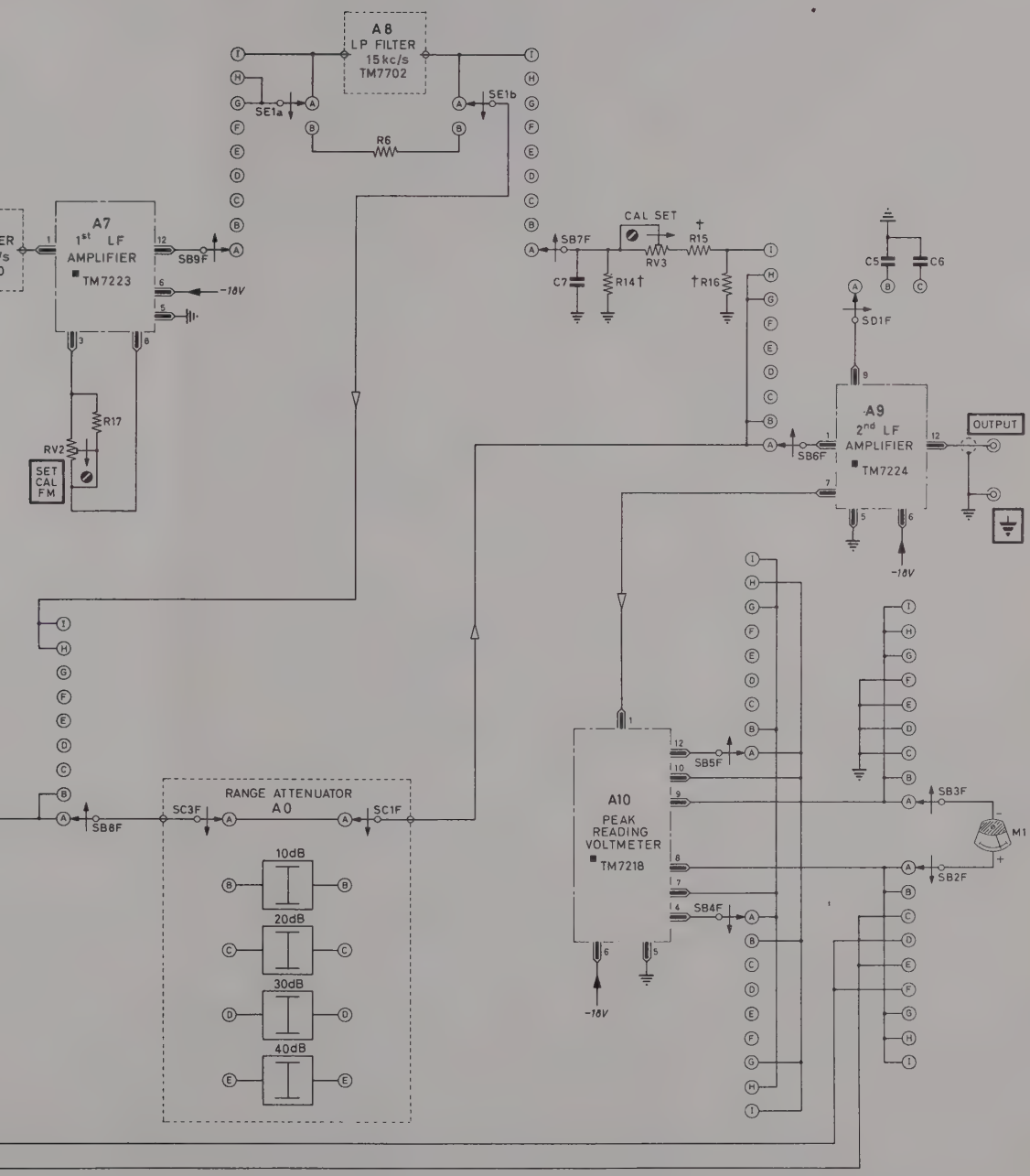
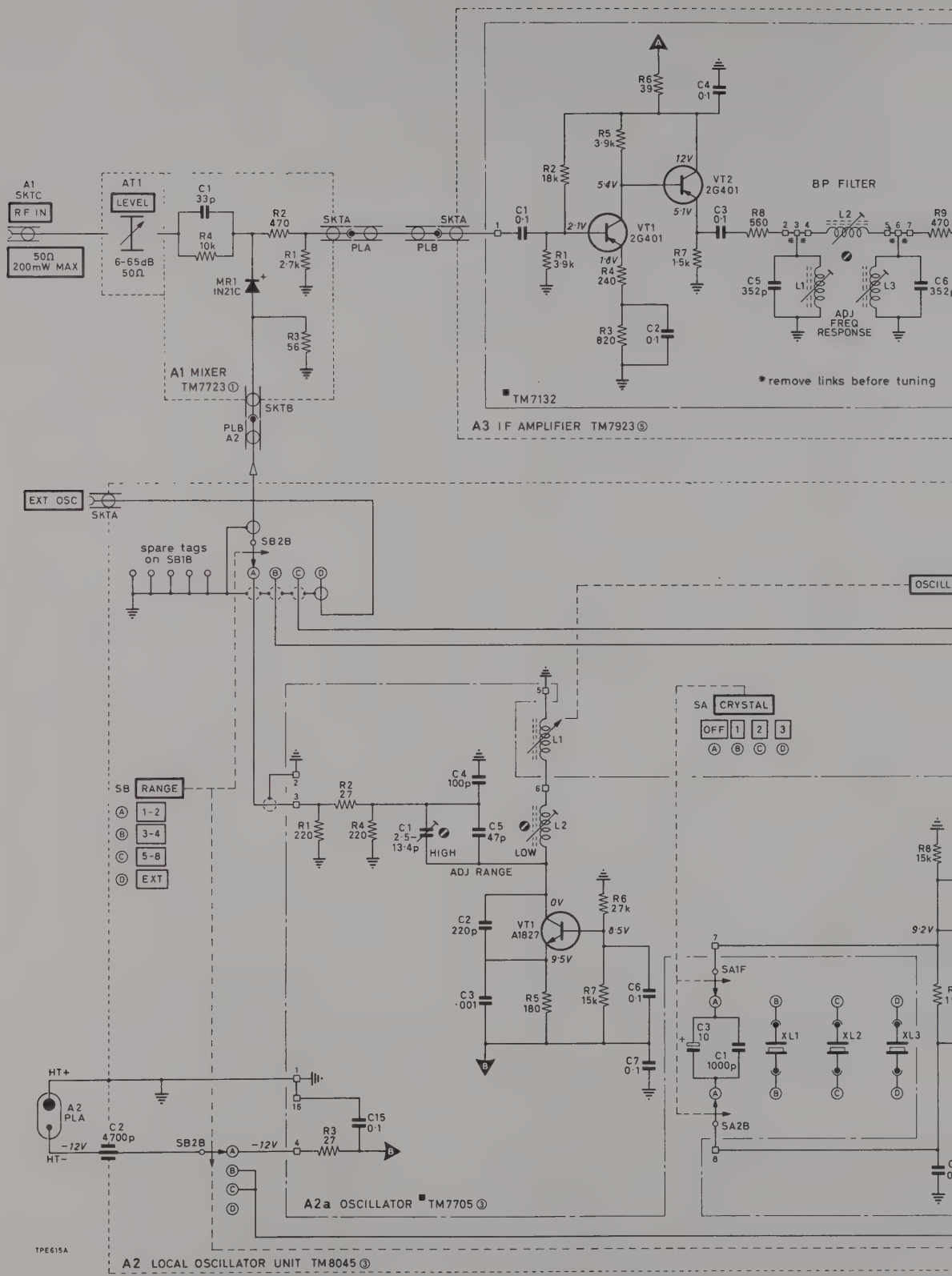


Fig. 7.1 Chassis inter-unit wiring


A1 SKTC
RF IN
50Ω 200mW MAX
AT1
LEVEL
6-65dB 50Ω
C1 33p
R4 10k
R2 470
R1 2·7k
SKTA
PLA
PLB
MR1 1N21C
R3 56
A1 MIXER TM7723 ①
SKTB
PLB A2
EXT OSC
SKTA
spare tags on SB1B
SB2B
A3 IF AMPLIFIER TM7923 ⑤
TM7132
R6 39
C4 0·1
R5 3·9k
R2 18k
12V
5·4V
VT2 2G401
5·1V
BP FILTER
C1 0·1
2·1V
VT1 2G401
C3 0·1
R8 560
L2
R9 470
R1 3·9k
1·6V
R4 240
R7 1·5k
C5 352p
L1
L3
C6 352p
ADJ FREQ RESPONSE
R3 820
C2 0·1
*remove links before tuning
SB RANGE
A 1-2
B 3-4
C 5-8
D EXT
R2 27
C4 100p
L1
R1 220
R4 220
C1 2·5-13·4p
HIGH
C5 47p
L2
LOW
ADJ RANGE
SA CRYSTAL
OFF 1 2 3
0V
R6 27k
C2 220p
VT1 A1827
8·5V
9·5V
C3 ·001
R5 180
R7 15k
C6 0·1
C7 0·1
SA1F
C3 10
C1 1000p
XL1
XL2
XL3
SA2B
R8 15k
9·2V
HT+
A2 PLA
HT-
-12V
C2 4700p
SB2B
C15 0·1
R3 27
A2a OSCILLATOR TM7705 ③
TPE615A
A2 LOCAL OSCILLATOR UNIT TM8045 ③

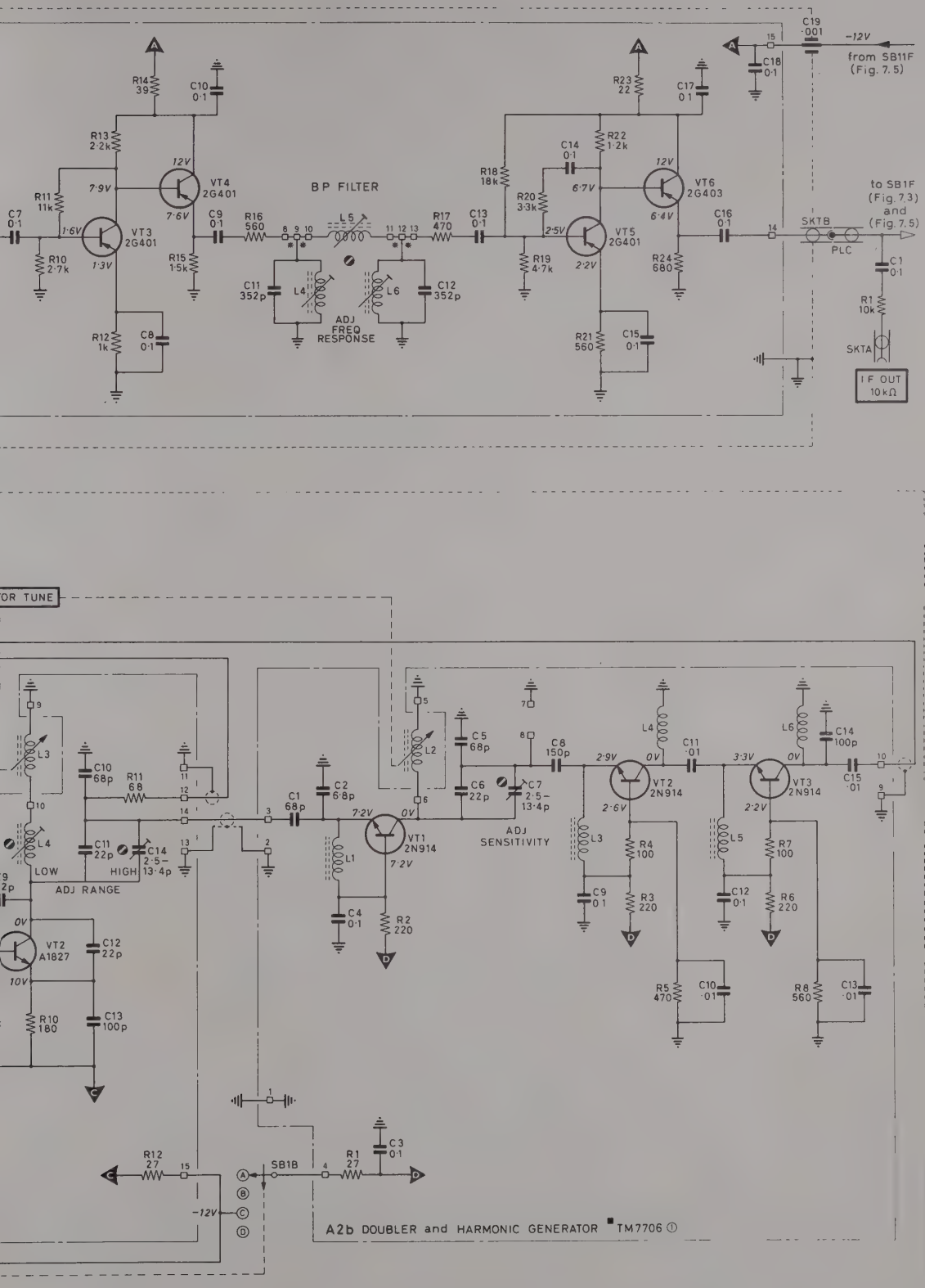
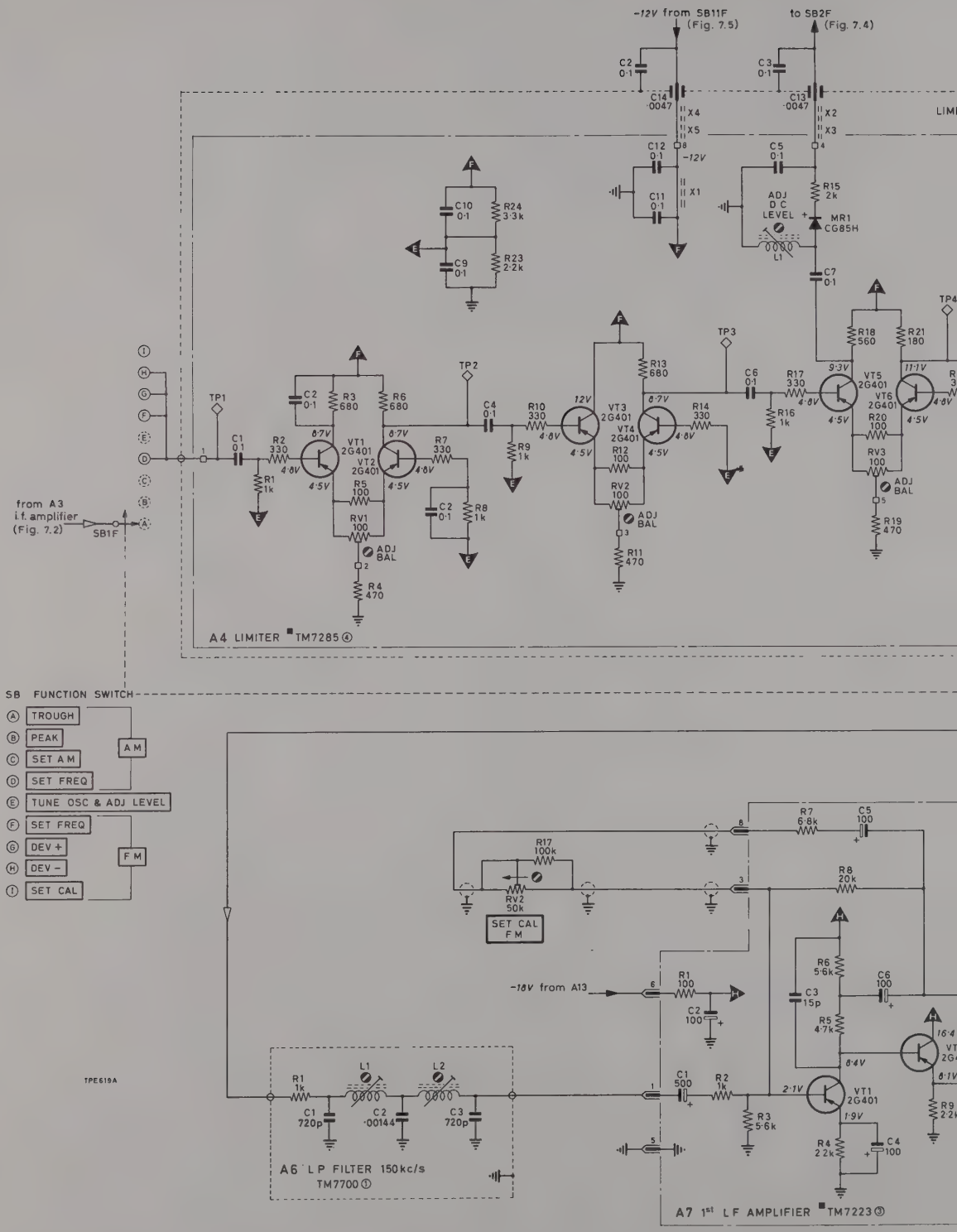


Fig. 7.2 Mixer, oscillator and i.f. amplifier


-12V from SB11F (Fig. 7.5)
to SB2F (Fig. 7.4)
A4 LIMITER TM7285 ④
from A3 i.f. amplifier (Fig. 7.2)
SB1F
SB FUNCTION SWITCH
A TROUGH
B PEAK
C SET AM
D SET FREQ
E TUNE OSC & ADJ LEVEL
F SET FREQ
G DEV +
H DEV −
I SET CAL
AM
FM
TPE619A
A6 LP FILTER 150kc/s TM7700 ①
SET CAL FM
-18V from A13
A7 1st LF AMPLIFIER TM7223 ③

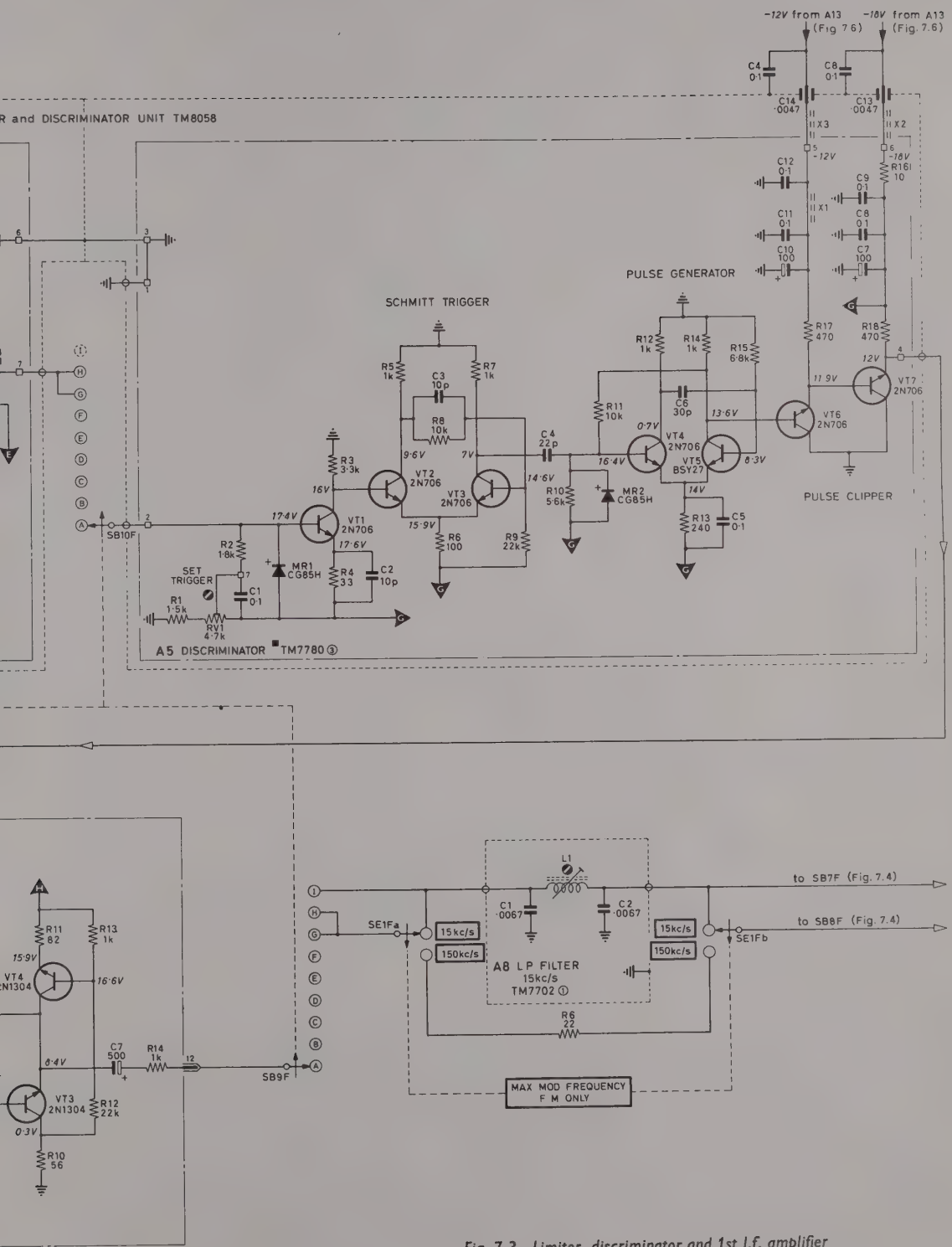
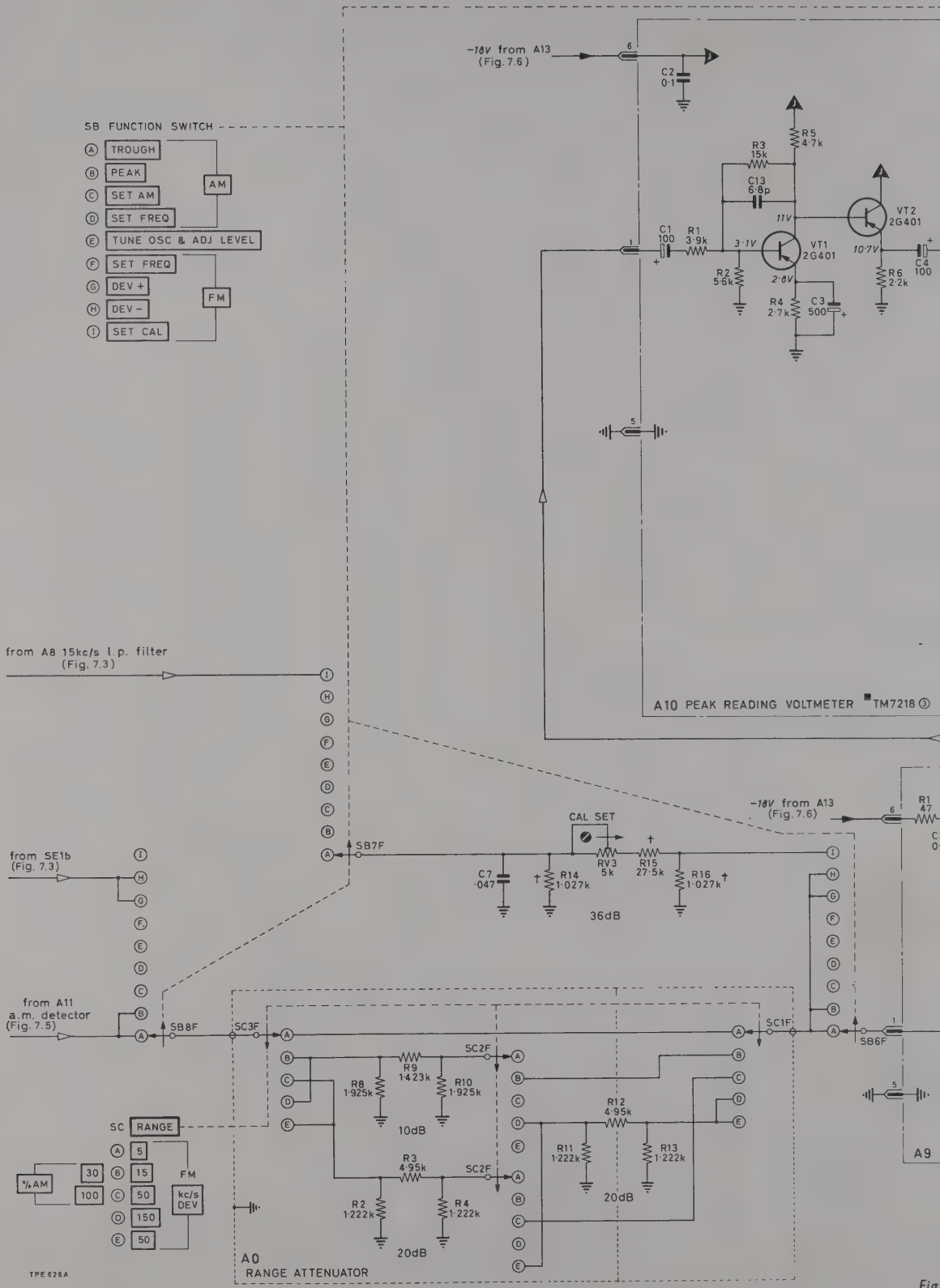
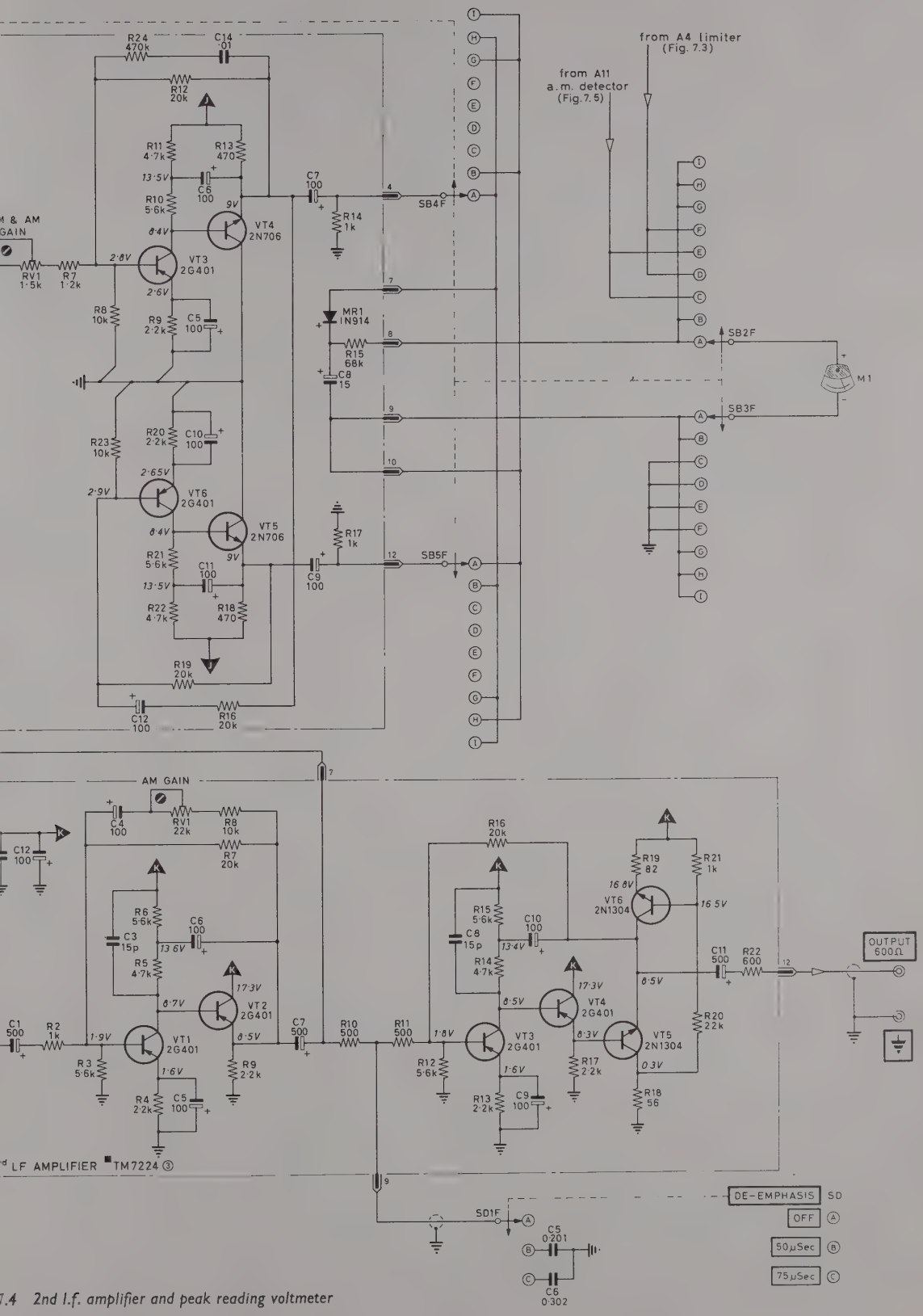


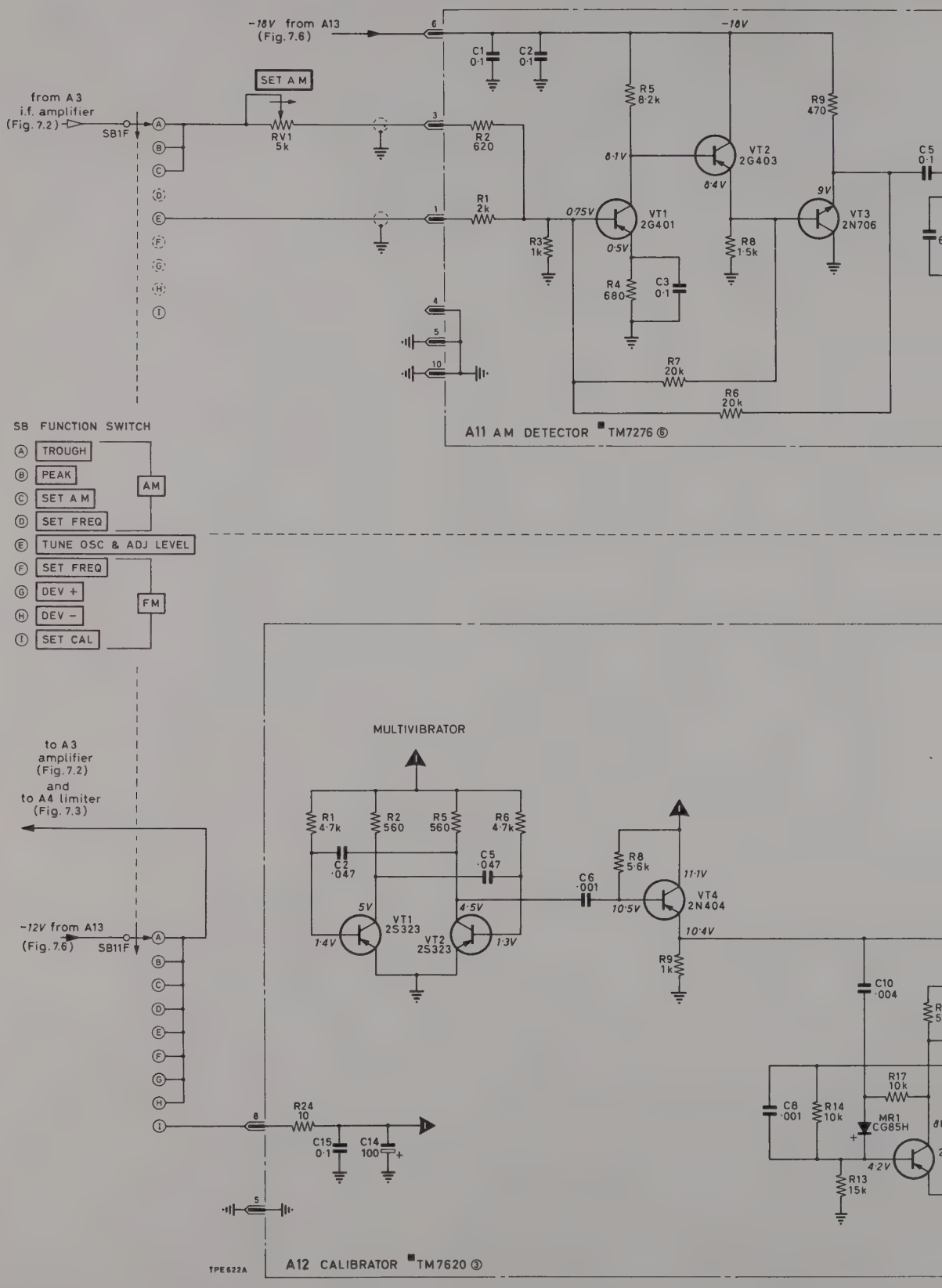
Fig. 7.3 Limiter, discriminator and 1st l.f. amplifier


SB FUNCTION SWITCH
A TROUGH
B PEAK
C SET AM
D SET FREQ
E TUNE OSC & ADJ LEVEL
F SET FREQ
G DEV +
H DEV −
I SET CAL
AM
FM
−18V from A13 (Fig. 7.6)
C2 0·1
R5 4·7k
R3 15k
C13 6·8p
VT2 2G401
11V
C1 100
R1 3·9k
3·1V
VT1 2G401
10·7V
C4 100
R2 5·6k
2·8V
R6 2·2k
R4 2·7k
C3 500
A10 PEAK READING VOLTMETER
TM7218 ③
from A8 15kc/s l.p. filter (Fig. 7.3)
−18V from A13 (Fig. 7.6)
R1 47
CAL SET
SB7F
RV3 5k
R15 27·5k
C7 ·047
R14 1·027k
R16 1·027k
36dB
from SE1b (Fig. 7.3)
from A11 a.m. detector (Fig. 7.5)
SB8F
SC3F
SC1F
SB6F
SC2F
R9 1·423k
R8 1·925k
R10 1·925k
10dB
R12 4·95k
R11 1·222k
R13 1·222k
20dB
R3 4·95k
R2 1·222k
R4 1·222k
20dB
SC RANGE
5
30
15
100
50
150
50
%AM
FM
kc/s DEV
A0
RANGE ATTENUATOR
A9
TPE 626A

Fig



7.4 2nd l.f. amplifier and peak reading voltmeter


-18V from A13 (Fig. 7.6)
from A3 i.f. amplifier (Fig. 7.2)
SB1F
SET A M
RV1 5k
R2 620
R1 2k
C1 0·1
C2 0·1
R5 8·2k
R9 470
VT2 2G403
VT1 2G401
VT3 2N706
R3 1k
R4 680
C3 0·1
R8 1·5k
R7 20k
R6 20k
C5 0·1
A11 AM DETECTOR TM7276 ⑤
SB FUNCTION SWITCH
TROUGH
PEAK
SET A M
SET FREQ
TUNE OSC & ADJ LEVEL
SET FREQ
DEV +
DEV −
SET CAL
AM
FM
to A3 amplifier (Fig. 7.2) and to A4 limiter (Fig. 7.3)
-12V from A13 (Fig. 7.6)
SB11F
MULTIVIBRATOR
R1 4·7k
R2 560
R5 560
R6 4·7k
C2 ·047
C5 ·047
VT1 2S323
VT2 2S323
C6 ·001
R8 5·6k
VT4 2N404
R9 1k
C10 ·004
R24 10
C15 0·1
C14 100
C8 ·001
R14 10k
R17 10k
MR1 CG85H
R13 15k
TPE622A
A12 CALIBRATOR TM7620 ③

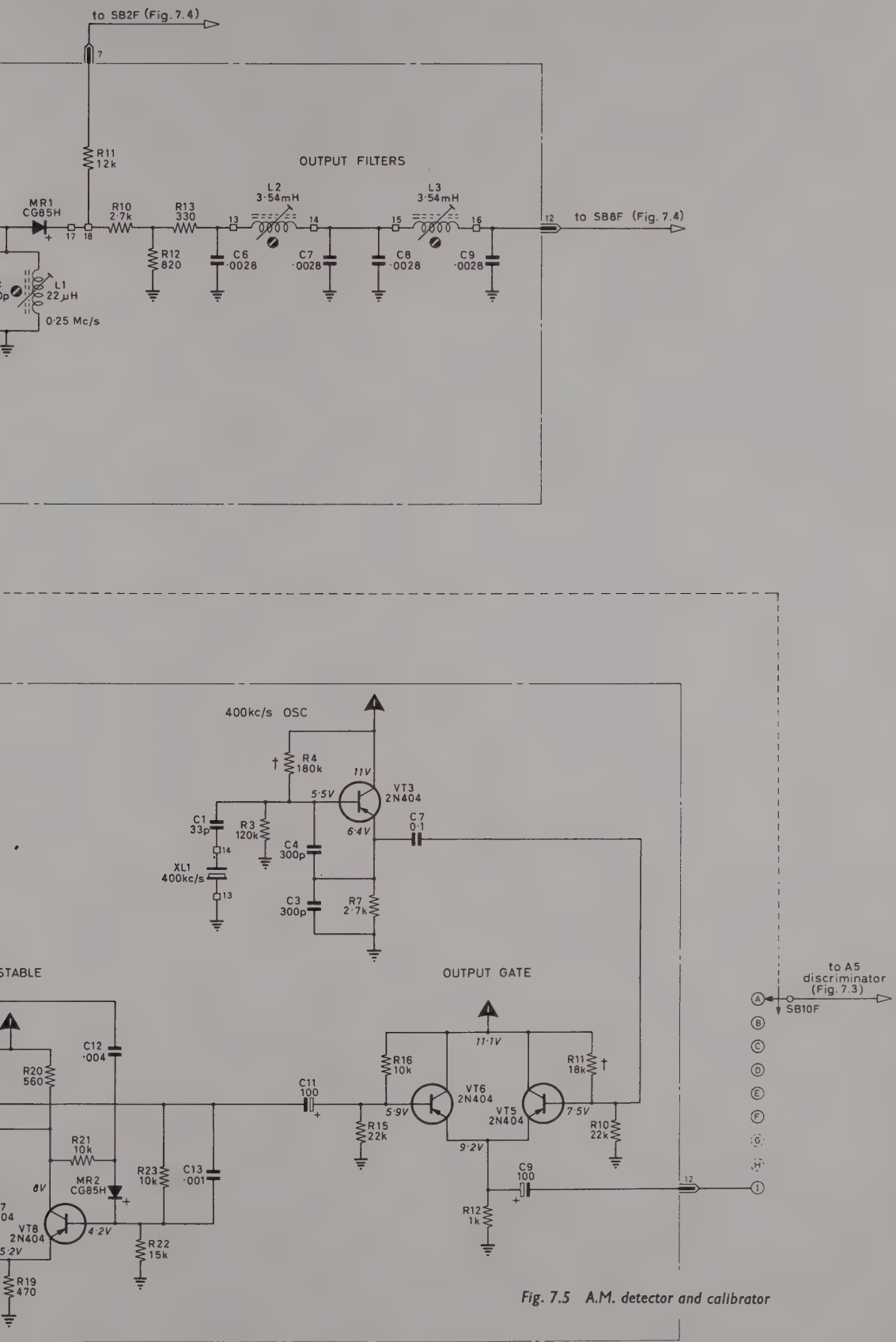
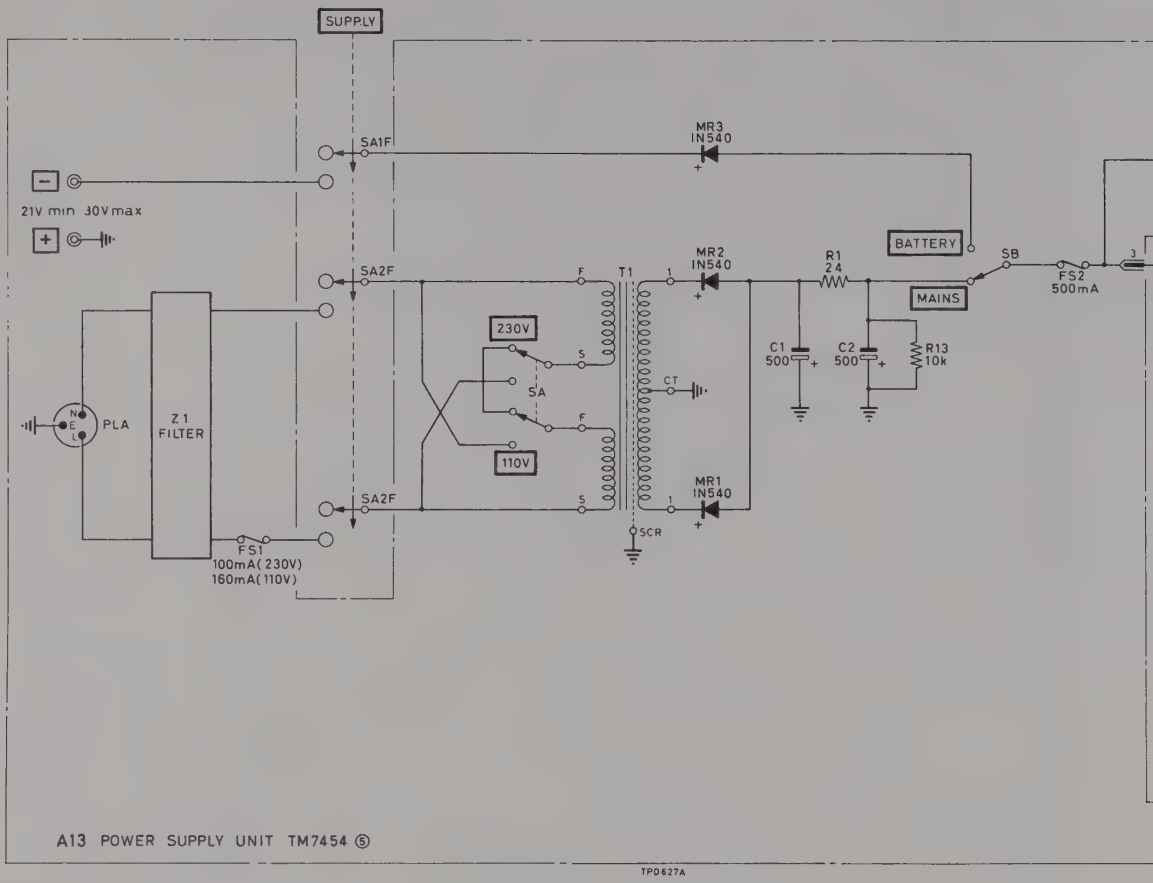


Fig. 7.5 A.M. detector and calibrator


SUPPLY
SA1F
MR3
IN540
21V min 30V max
BATTERY
SB
MR2
IN540
R1
24
SA2F
F
T1
1
MAINS
FS2
500mA
230V
S
C1
500
C2
500
R13
10k
CT
SA
F
110V
MR1
IN540
SA2F
S
1
SCR
N
E
L
PLA
Z1
FILTER
FS1
100mA(230V)
160mA(110V)
A13 POWER SUPPLY UNIT TM7454 ⑤
TPD627A

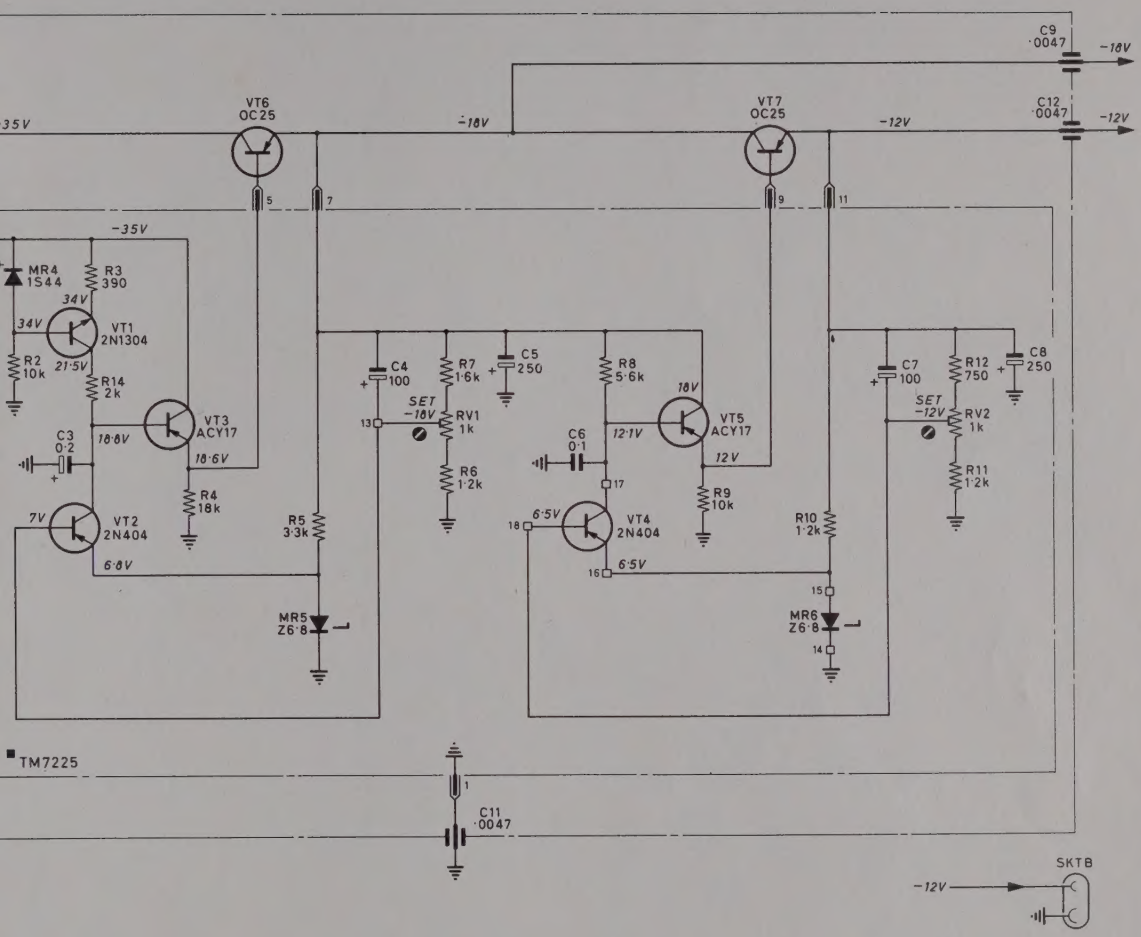


Fig. 7.6 Power supply unit

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